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# **Errata**

Title & Document Type: 6824A Power Supply / Amplifier Operating and Service

Manual

Manual Part Number: 06824-90001

**Revision Date: July 1966** 

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POWER SUPPLY/AMPLIFIER PS/A SERIES, MODEL 6824A SERIAL NUMBER PREFIX 6F

Printed: July, 1966

Stock Number: 06824-90001

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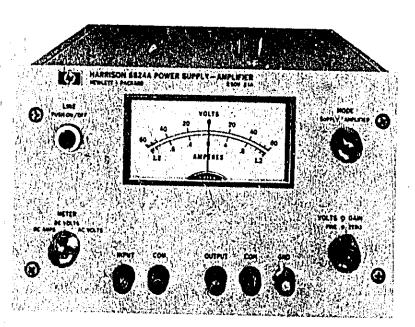


Figure 1-1. DC Power Supply/Amplifier, Model 6824A

# SECTION I GENERAL INFORMATION

# 1-1 DESCRIPTION

The Power Supply/Amplifier (P3/A) shown on Figure 1-1, is a general purpose instrument useful in any laboratory engaged in research and development of electronic systems or components. The PS/A can be operated in one of two basic operating modes; power supply or amplifier. Terminals on the rear barrier strip permit access to various control points within the unit to further expand the operating capabilities of the instrument. The resulting flexibility lends the PS/A to an almost unlimited number of applications. Some of these applications are outlined in Section III of this manual; but a more comprehensive description of the features and applications of the PS/A are included in Application Note 82 published by the Harrison Division, A copy of Application Note 82 can be obtained from your local Hewlett-Packard field office. The following paragraphs describe some of the features of the PS/A both as a power supply and an amplifier.

# 1-3 POWER SUPPLY FEATURES

- 1-4 The unit can be made to function as a regulated dc power supply by setting the front panel MODE switch to the SUPPLY position. The supply can furnish either a Constant Voltage output or a Constant Current output (with the addition of an external current sampling resistor). The dc output is bi-polar and is continuously adjustable from its maximum rated positive value to an equal negative value; smoothly through zero with no polarity switch. Both the supply and the load are protected against overloads by a fixed current limit which is set by means of an internal adjustment.
- 1-5 The supply can be programmed (controlled) at a very high rate of speed (less than 50 µsec for output voltage change over the entire voltage span). The supply can be programmed locally; by means of the front panel control, or remotely; by means of a resistance, voltage, or current source.

1-6 A single meter is used to measure dc output voltage or current. A front panel METER switch allows the dual purpose selection.

#### 1-7 AMPLIFIER FEATURES

- 1-8 As a power amplifier the unit has a high signal-to-noise ratio (80 db at full output) and an adjustable gain of from 0 to 10 (20 db). The output distortion is low less than 0.02% at i KHz. The amplifier contains a push-pull output stage that can can furnish either a Constant Voltage output or a Constant Current output (with the addition of an external current sampling resistor). The external input can be from either a voltage or a current source. The bandwidth of the amplifier is from dc to 20KHz (±3db).
- 1-9 The ac component of the output signal can be measured with the front panel meter (Model 6824A only).

# 1-10 SPECIFICATIONS

1-11 Detailed specification for the PS/A are given in Table 1-1.

# 1-12 OPTIONS

1-13 Options are factory modifications of a standard instrument that are requested by the customer. The following option is available for the instrument covered by this manual.

# Option No. Description

28

Rewire For 230V AC Input: Supply as normally shipped is wired for 115 Vac input. Option 28 consists of reconnecting the input transformer for 230 Vac operation.

# 1-14 ACCESSORIES

1-15 The applicable accessories listed in the following chart may be ordered with the instrument or

separately from your local Hewlett-Packard field sales office (refer to list at rear of manual for addresses).

Part No.	Description
14513A	Rack Kit for mounting one 3½" -high unit. (Refer to Section II for details.)
14523A	Rack Kit for mounting two $3\frac{1}{2}$ " -high units. (Refer to Section II for details.)
14515A	Rack Kit for mounting one 5½" -high unit. (Refer to Section II for details.)
14525A	Rack Kit for mounting two 51" -high units. (Refer to Section II for details)

# 1-16 INSTRUMENT IDENTIFICATION

1-17 Hewlett-Packard instruments are identified by a three-part serial number tag. The first part is

the unit model number. The second part is the serial number prefix, which consists of a number-letter combination that denotes the date of a significant design change. The number designates the year, and the letter A through L designates the month, January through December respectively. The third part is the instrument serial number.

1-18 If the serial number prefix on your unit does not agree with the prefix on the title page of this manual, change sheets are included to update the manual. Where applicable, backdating information is given in an appendix at the rear of the manual.

# 1-19 ORDERING ADDITIONAL MANUALS

1-20 One manual is shipped with each instrument, Additional manuals may be purchased from your local Hewlett-Packard field office (see list at rear of this manual for addresses). Specify the model number, serial number prefix, and  $\mathfrak P$  stock number provided on the title page.

#### DC POWER SUPPLY POWER AMPLIFIER OUTPUT: OUTPUT: -50 to +50 Vdc @ 0 to 1.0 amp. 100 volts peak-peak @ 0 to 1.0 amp. LOAD REGULATION: VOLTAGE GAIN: Less than 0,02% plus 5 mV for a full load Gain is variable, 0 to 10 (20db). Output is to no load, change in output current. inverted. LINE REGULATION: Less than 0,02% plus 5 mV for any line FREQUENCY RESPONSE: voltage change within the input rating, DC to 20 kHz (±3db) at full output. RIPPLE AND NOISE: Less than 10 mVrms. DISTORTION: REMOTE PROGRAMMING: Remote programming of the output at ap-Output distortion is less than 0.02% at proximately 500 ohms per volt is made available 1 kHz and maximum rated output. at the rear terminals. INPUT IMPEDANCE: PROGRAMMING SPEED: Approximately 2K ohms. Less than 50 usec are required to program between -50V and +50V. Typically, about 15 usec MAXIMUM PHASE SHIFT: is required to program between 10% and 90% of the maximum voltage span. At dc -180° TRANSIENT RECOVERY TIME: At 100 Hz - 180, 70

INPUT: 105-125/210-250 Vac, single phase, 50-60 Hz; 1, 3 Amp, 96 Watts maximum.

STABILITY: Less than 0, 075% +5 mV total drift for 8 hours after an initial warm-up of 30 minutes at constant line voltage with ambient temerature variations held to  $\pm 3^{\circ}$ C.

At 10 kHz - 2050

At 20 kHz - 2250

111

TEMPERATURE RANGES: Operating: 0 to 50°C. Storage: -20 to +85°C.

Less than 100 usec for output recovery to

within 5 mV plus 0, 02% following a full load

current change in the output.

TEMPERATURE COEFFICIENT: Less than 0.015% plus 1 mV per degree Centigrade.

OUTPUT IMPEDANCE: Less than 0.03 ohm from dc to 100 Hz. Less than 0.3 ohm from 100 to 1000 Hz. Less than 3 ohms from 1000 Hz to 100 kHz.

OVERLOAD PROTECTION: A continuously acting current limiting circuit protects the unit for all overloads including a direct short placed across the output terminals.

METER: The front panel meter is triple purpose. It can be used as a -60 Vdc to +60 Vdc voltmeter, a -1, 2 Amp to +1, 2 Amp dc Ammeter, or as a 0 to 60 Vac rms voltmeter.

SIZE:  $5\frac{1}{4}$ " H x  $8\frac{1}{4}$ " W x 14 5/8" D. Two units can be rack mounted side-by-side in a standard rack panel.

WEIGHT: 17 lbs. nct, 21 lbs. shipping.

FINISH: Light gray front panel with dark gray case.

# SECTION II INSTALLATION

# 2-1 INITIAL INSPECTION

2-2 Before shipment, this instrument was inspected and found to be free of mechanical and electrical defects. As soon as the instrument is unpacked, inspect for any damage that may have occurred in transit. Save all packing materials until the inspection is completed. If damage is found, proceed as described in the Claim for Damage in Shipment section of the warranty page at the rear of this manual.

# 2-3 MECHANICAL CHECK

2-4 This check should confirm that there are no broken knobs or connectors, that the cabinet and panel surfaces are free of dents and scratches, and that the meter is not scratched or cracked.

# 2-5 ELECTRICAL CHECK

2-6 The instrument should be checked against its electrical specifications. Section V includes

an "in-cabinet" performance check to verify power instrument operation.

# 2-7 <u>INSTALLATION DATA</u>

2-8 The instrument is shipped ready for bench operation. It is necessary only to connect the instrument to a source of power and it is ready for operation.

#### 2-9 LOCATION

2-10 This instrument is cooled with forced air from the cooling fan. However, it should not be used in an area where the ambient temperature may exceed 50°C.

# 2-11 RACK MOUNTING

2-12 This instrument may be rack mounted in a standard 19 inch rack panel either alongside a similar unit or by itself. Figures 2-1 and 2-2 show how both types of installations are accomplished.

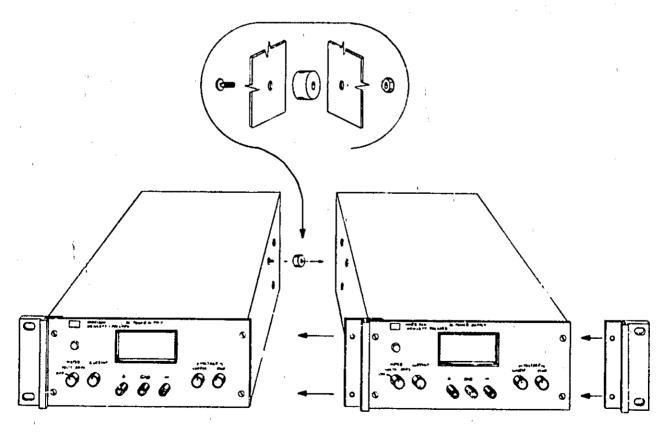


Figure 2-1. Rack Mounting, Two Units

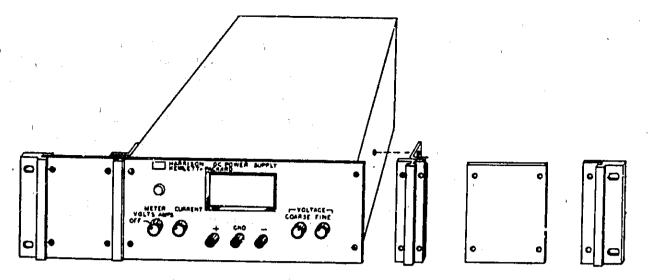


Figure 2-2. Rack Mounting, One Unit

- 2-13 To mount two units side-by-side, proceed as follows:
- a. Remove the four screws from the front panels of both units.
- b. Slide rack mounting ears between the front panel and case of each unit.
- c. Slide combining strip between the front panels and cases of the two units.
- d. After fastening rear portions of units together using the bolt, nut, and spacer, replace panel screws.
- 2-14 To mount a single unit in the rack panel proceed as follows:
- a. Bolt rack mounting ears, combining straps, and angle brackets to each side of center spacing panels. Angle brackets are placed behind combining straps as shown in Figure 2-2.
- b. Remove four screws from front panel of unit.
- c. Slide combining strips between front panel and case of unit.
- d. Bolt angle brackets to front sides of case and replace front panel screws.

# 2-15 INPUT POWER REQUIREMENTS

2-16 This power supply may be operated from aither a nominal 115 volt or 230 volt 59-60 cycle power source. The unit, as shipped from the factory, is wired for 115 volt operation. The input power required when operated from a 115 volt 60 cycle power source at full load is 96 watts and 1.3 amperes.

# 2-17 CONNECTIONS FOR 230 VOLT OPERATION (Figure 2-3)

2-18 Normally, the two primary windings of the input transformer are connected in parallel for operation from 115 volt source. To convert the

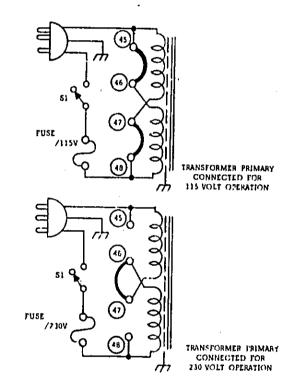


Figure 2-3. Frimary Connections

power supply to operation from a 230 volt source, the power transformer windings are connected in series as follows:

- a. Unplug the line cord and remove the top and bottom covers from the unit.
- b. Break the printed wiring between 45 and 46 and also between 47 and 48 on the printed circuit board. These are shown in Figure 2-3, and are labeled on printed side of the circuit board.
  - c. Connect a strap between 46 and 47.
- d. Replace existing fuse with 1.5 ampere, 230 volt fuse. Replace covers and operate unit normally.

# 2-19 POWER CABLE

2-20 To protect operating personnel, the National Electrical Manufacturers Association (NEMA) recommends that the instrument panel and cabinet be grounded. This instrument is equipped with a three conductor power cable. The third conductor is the ground conductor and when the cable is plugged into an appropriate receptable, the instrument is grounded. The offset pin on the power cable three-prong connector is the ground connection.

2-21 To preceive the protection feature when operating the instrument from a two-contact outlet, use a three-prong to two-prong adapter and connect the green lead on the adapter to ground.

# 2-22 REPACKAGING FOR SHIPMENT

2-23 To insure safe shipment of the instrument, it is recommended that the package designed for the instrument be used. The original packaging material is reusable. If it is not available, contact your local Hewlett-Packard field office to obtain the materials. This office will also furnish the address of the nearest service office to which the instrument can be shipped. Be sure to attach a tag to the instrument which specifies the owner, model number, full serial number, and service required, or a brief description of the trouble.

1

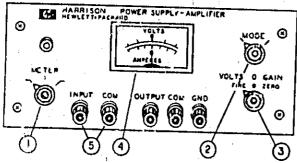
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# SECTION III OPERATING INSTRUCTIONS

#### 3-1 OPERATING CONTROLS AND INDICATORS

3-2 The front panel controls and indicators, together with the normal turn-on sequences, for both power supply and amplifier operation are shown on Figure 3-1. These turn-on procedures are performed utilizing the normal rear terminal strapping connections as received from the factory.



TURN-ON SEQUENCE - POWER SUPPLY

- 1. SET AC LINE SWITCH TO ON AND OBSERVE THAT PILOT LIGHT GOE! ON
- SET MATER SWITCH (1) TO DC VOLTS.
- SET MODE SWITCH (2) TO SUPPLY.
- AD JST VOLTS/GAIN CONTROL (3) CLOCKWISE OR COUNTERGLOCK-WISE UNTIL DESIRED POSITIVE, OR NEGATIVE OUTPUT VOLTAGE, RESPECTIVELY, IS INDICATED ON FRONT PANEL METER (4). CONTROL OF THE OUTPUT VOLTAGE. THE OUTPUT VOLTAGE,
- SHORT CIRCUIT OUTPUT TERMINALS: SET METER SWITCH TO AMPS AND OBSERVE SHORT CIRCUIT OUTPUT CURPENT ON METER. CURRENT LIMIT ID FACTORY ADJUSTED SO THAT METER SHOULD INDICATE FULL SCALE DEFLECTION:
- REMOVE SHORT AND CONNECT LOAD TO OUTPUT TERMINALS

# TJAN-ON SEQUENCE - AMPLIFIER

- TURN ON UNIT AND SET MCDE SWITCH (2) TO AMPLIFIER. 2,
- CONNECT INPUT SIGNAL SOURCE TO INPUT (5) TERMINALS (FRONT OR REAR)
- ADJUST VOLTS/GAIN CONTROL (1) FOR DESIRED GAIN (0 TO 10).
  CONCENTRIC FINE/ZERO CONTROL PERMITS AVERAGE OUTPUT TO BE ADJUSTED TO ZERO.
- MEASURE OUTPUT VOLTAGE WITH EXTERNAL VOLTMETER FO 6823A UNITS OR FOR 6824A UNITS, SET METER SWITCH (1)
  TO AC VOLTS AND READ OUTPUT DIRECTLY ON FRONT PANEL
- 5. CONNECT LOAD TO OUTPUT TERMINALS (FRONT OR REAR),

Figure 3-1. Front Panel Controls and Indicators

# OPERATING MODES

The position of the front panel MODE switch determines whether the instrument will be used as a power supply or an amplifier. Terminals on the

rear barrier strip, which connect into various control points within the unit, allow strapping connections to be made which enable the power supply or amplifier to be utilized in an almost unlimited number of applications. The following paragraphs describe the procedures for utilizing some of the features which will suggest many related applications.

- 3-5 However, the configurations shown in the succeeding paragraphs by no means exhaust the various modes in which the PS/A can be used. The reader is encouraged to consider this instrument in terms of operational amplifier techniques; this viewhoc.  $\iota$  will suggest and facilitate many other areas of application. Many such possibilities are suggested by handbooks on operational amplifiers (e.g., "Handbook of Operational Amplifier Applicutions" - Burr-Brown Research Corporation, Copyright 1963).
- 3-6 Such handbooks suggest how operational amplifiers can be used as integraters, differentiaters, ramp function generators, etc.; with the PS/A, these techniques are extended to previously unattained power levels.
- An alternate viewpoint is to consider the PS/A as a general-purpose DC coupled precision power transducer capable of accepting either resistance, voltage, or current input (AC and/orDC) and delivering a power output, Constant Voltage or Constant Current, AC and/or DC.
- 3-8 Application Note 82, published by the Harrison Division, presents a more detailed approach concerning the operational features and related applications of the PS/A unit. A copy of this note can be obtained from your local Hewlett-Packard field office.

# 3-9 GENERAL NOTES AND OPERATING CONSID-**ERATIONS**

#### 3-10 DIAGRAMS

All of the following diagrams show the necessary barrier strip connections, as well as a simplified circuit diagram showing how the major internal components of the unit are connected for that strapping arrangement. The diagrams also

include all components which must be added externally to the unit. The barrier strip portion of each diagram is correct for all PS/A Series instruments, but when implementing these strapping patterns, check whether the terminal locations are the same as shown in the following diagrams, which are based specifically on the Model 6823A. In any case, the terminal designations are correct for all PS/A Series instruments and should be followed regardless of their location.

3-12 Unless otherwise indicated, the configurations shown have been selected so that operation is independent of whether the front panel MODE switch is in the power SUPPLY or AMPLIFIER position.

# 3-13 EXTERNAL RESISTORS

3-14 External resistors and potentiometers should be wire-wound and have a temperature coefficient of less than 20ppm/OC. They should operate at less than 1/30th (preferably 1/100th) of their wattage rating in order to minimize short term "bobble" associated with elevated surface temperature.

# 3-15 CONNECTING LEADS

3-16 Shielded leads should be used for all barrier strip connections, except that twisted leads may be used from the "Common" and "Output" terminals to the load. The outer sheath of the shielded wire should never be used as a conductor, but should be connected at the unit end only to the terminal designated "Common".

3-17 Best high speed programming and wide band amplifier operation will be achieved when the load is connected at the rear terminals rather than at the front. Any capacitance in parallel with the load device should be minimized, since this capacitance will limit the high frequency bandwidth.

3-18 Ensure that the screws on the rear terminals are tightened securely before power is applied to the unit.

# 3-19 FINE/ZERO CONTROL

3-20 The front panel Fine/Zero control is operable for all modes. In the power supply mode, it acts as a fine control on the DC output. In remote programming and amplifier modes, it provides a DC "offset" adjustment of approximately 1% of the maximum rated output voltage. Thus, this control permits the accurate setting of zero output voltage with zero ohms programming.

# 3-21 GROUND LOOPS

3-22 If excessive  $60~H_Z$  component exists on output, check for ground loops. The system in which the PS/A is employed should have one ground point only. See the Harrison Division's power supply Application Manual, Sections Dla (6) and Dld, for further discussion of typical ground loop problems.

# 3-23 REVERSE GURRENT LOADING

3-24 An active load connected to the unit may actually deliver a reverse current during a portion of it's operating cycle. An external source cannot be allowed to pump current into the instrument without the risk of possible damage. To avoid damage it is necessary to preload the unit with a dummy load resistor so that it will deliver current through the entire operating cycle of the load device.

3-25 In general, the load resistor that is added should draw a current equal to the peak current which otherwise would be forced back into the PS/A. This means that the current rating of the PS/A instrument used must be equal to, or greater than, the sums of the absolute values of the output and reverse currents.

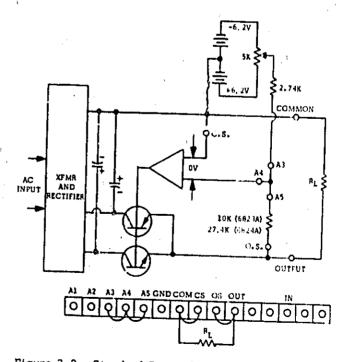


Figure 3-2. Standard Power Supply Operation (CV)

# 3-25 CONSTANT VOLTAGE OPERATING MODES

# 3-27 NORMAL OPERATING MODES

3-28 Figure 3-2 shows the circuit configuration and normal rear barrier strip strapping pattern for standard power supply operation. The front panel MODE switch is in the SUPPLY position. The DC output is continuously adjustable from its maximum rated positive value to an equal negative value — smoothly through zero with no polarity switch.

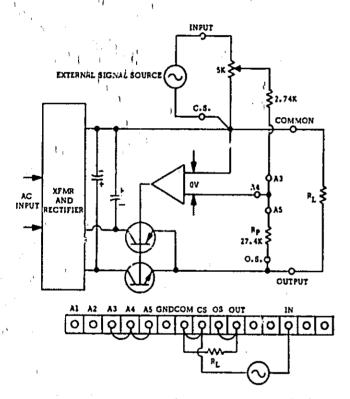


Figure 3-3. Standard Amplifier Operation (CV)

3-29 Similarly, Figure 3-3 shows the circuit configuration and strapping pattern for standard amplifier operation. The external oscillator must have a voltage output which is at least 1/10th of the peak output voltage desired; its current capability must be at least 600 µa for each volt input - in other words, the oscillator faces an input impedance of approximately 1.7K. The front panel gain control permits the amplifier gain to be set anywhere from zero to X10, although the gain setting will not be a linear function of control rotation. The front panel FINE/ZERO control permits the dc component of the output to be adjusted approximately 1% of the voltage rating of the instrument, thus enabling the average output value to be adjusted exactly to zero (assuming the input signal has either no dc component or only a small dc component). If desired, a coupling

capacitor can be added to block any large input dc components, although in many applications this capacitor will be omitted so that the PS/A can amplify <u>all</u> input signal components, both dc and ac.

#### 3-30 REMOTE SENSING

3-31 In normal operation, a power supply develops an IR drop in the leads connecting it to the load. This reduces the output voltage actually present at the load terminals and also degrades the load regulation performance.

3-32 Sensing terminals permit the feedback amplifier to sense at the load terminals instead of at the PS/A output terminals. This configuration is shown in Figure 3-4 and is applicable for all Constant Voltage modes of operation, both power supply and amplifier. Moreover, the strapping pattern shown in Figure 3-4 can be combined with any of the Constant Voltage diagrams shown elsewhere in the manual in order to achieve optimum performance at the load terminals.

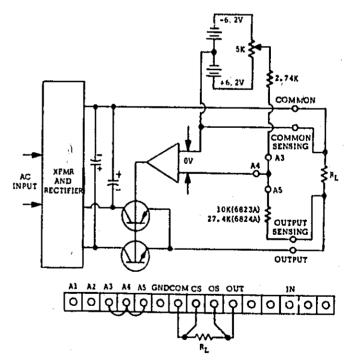


Figure 3-4. Remote Error Sensing

3-33 This does not mean that remote sensing can be used to compensate for an unlimited IR drop between the Common and Output terminals and the load. Any voltage lost in this way detracts directly from the output voltage available for the load. Furthermore, proper operation of the PS/A is insured only if the IR drop in the Common lead is less than, or equal to, one volt.

3-34 If external switching schemes are employed in conjunction with remote sensing, care must be taken that each sensing lead is not momentarily open circuited with respect to the corresponding output terminal. Such open circuit conditions can result in the power supply delivering excessive voltage and/or current to the load, risking damage to both the PS/A and the load device. If it is impossible to insure that an open sensing lead condition will not occur, it is recommended that a 100 ohm, 1/2 watt resistor be connected from "Common" to "Common Sensing", and a similar resistor fron "Output" to "Output Sensing". This will minimize voltage and current transients if the sensing leads should accidentally become open circuited. All sensing leads should be shielded.

# 3-35 REMOTE PROGRAMMING, POSITIVE OUTPUT

3-36 Figure 3-5 shows the strapping pattern and associated circuit configuration for remote programming Power Supply / Amplifiers for positive Constant Voltage DC output at the rate of 500 ohms/volt. The programming control Rp should be wire-wound and selected in accordance with paragraph 3-13. Its dissipation can be computed by remembering that the current through Rp is 2 MA.

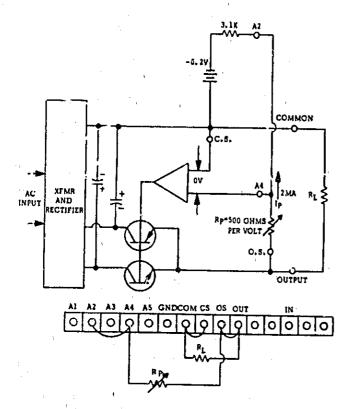


Figure 3-5. Remote Resistance Programming,
Positive Output

3-37 Figure 3-6 shows two resistance programming characteristics which are typical (although somewhat exaggerated) of the performance without programming alignment. The output voltage with zero programming always will differ from zero, being either positive (1) or negative (A), and the maximum output voltage will be achieved with a value of programming resis ance not predicted by the programming coefficient.

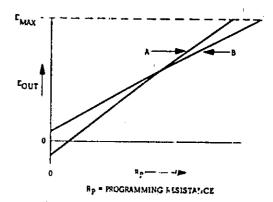


Figure 3-6. Unaligned Resistance Programming Characteristics

3-38 Notice, however, that even with an unaligned instrument the typical characteristics of A and B are extremely linear — that is, there is no curvature. To convert either of the characteristics of Figure 3-6 to the desired programming characteristic of Figure 3-7, it is necessary to adjust both the zero crossing and the angle of slope of the programming characteristic.

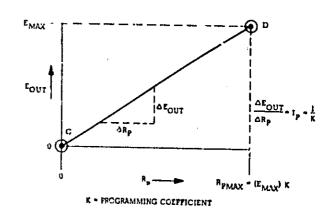


Figure 3-7. Aligned Resistance Programming
Transfer Function

3-39 In the PS/A Series the zero crossing is adjusted by means of the front panel FINE/ZERO knob, while the slope of the programming function

is adjusted by means of an internal control, R38. First the zero crossing is adjusted; then a programming resistance, which according to the programming coefficient (500 ohms/volt) should yield the maximum rated output voltage for the instrument, is connected to the programming terminals, and R38 is adjusted until the output voltage is exactly equal to the maximum rated value. Since the two adjustments are mildly interdependent, it is necessary to repeat this rocedure once.

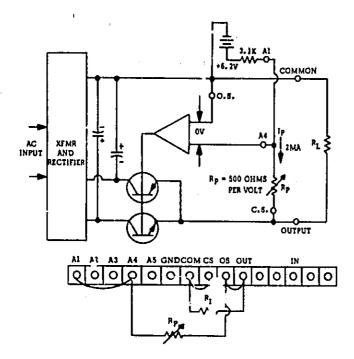


Figure 3-8. Remote Resistance Programming, Negative Output

# 3-40 REMOTE PROGRAMMING, NEGATIVE OUTPUT

3-41 Figure 3-8 is the same as Figure 3-5 except that the linearly programmed output voltage is negative. The internal programming slope adjustment for a negative output is R39.

# 3-42 REMOTE PROGRAMMING, BIPOLAR OUTPUT

3-43 The circuits of Figures 3-5 and 3-8 can be combined to yield a power supply which is programmable in either direction, depending upon the position of switch S1 (see Figure 3-9). One obtains a linearly programmed negative DC output when switch S1 is connected to A1; and a positive output when connected to A2. Switch S1 must be a break-before-make switch.

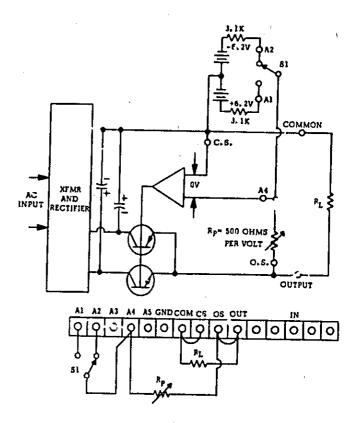


Figure 3-9. Remote Resistance Programming, Bi-Polar Output (CV)

3-44 The advantage of the configuration of Figure 3-9 is that the output is <u>linearly</u> programmed in both directions. The disadvantage is that a switch must be used in order to reverse the output polarity.

3-45 If it is desired to achiev accurate programming reproducibility, or if it is pecessary to repeatedly set the power supply output to specific voltage values, either of the substitute arrangements suggested by Figure 3-10(B) and 3-10(C)may be employed in place of the rheostat Rp. Suppose, for example, it is decided to set the power supply over the span from 0-20V. If continuous adjustment is required, the best choice would be a 10K rheostat (since the programming coefficient is 500 ohms/volt). But if specific tests must be made at 0, 5, 10, 15, and 20V, the arrangement shown in Figure 3-10(B) should be employed, using a make-before-break switch. If the same voltage values are desired, but not in ascending or descending sequence, the terminals from the resistor string can be connected to the switch in any desired sequence, as suggested by Figure 3-10(C).

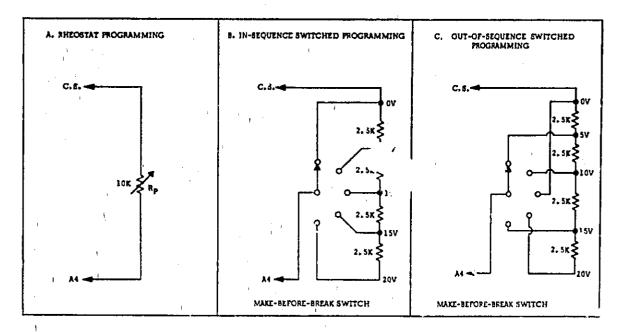


Figure 3-10. Switched Resistor Substitutes

3-46 These substitute configurations for Rp can be employed with any of the diagrams in this section. In all such applications a make-before-break switch must be used so that output voltage transients will not occur during the switching interval. Such transients can lead to damage to both the load and the power supply.

# 3-47 SWITCHLESS BIPOLAR PROGRAMMING

3-48 Figure 3-11 shows how to resistance program a PS/A for continuously variable output do voltage, remotely controlled, without need for a polarity switch. Clockwise rotation of the 10K potentiometer will result in a positive output equal to the maximum voltage capability of the instrument.

3-49 The advantage of Figure 3-11 is that the output is continuously variable through zero with one control. The disadvantage is that the output is not a linear function of the rotation of potentiometer RQ. Note, however, that in many applications it is not necessary that the programming function be linear, just reproducible. The power supply can be switched readily to positive or negative values (and later reset precisely at these same values) by substituting a switched resistor network for RQ. The switch must be a <a href="mailto:break-before-make">break-before-make</a> type.

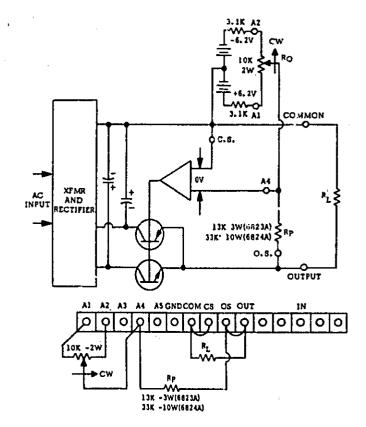


Figure 3-11. Bi-Polar Programming (Constant Voltage)

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<u>.</u>

3-51 In some applications it is desirable to add an AC component to the adjustable DC output of a Constant Voltage power supply. Figure 3-12 shows how this can be done.

3-52 The PS/A is operated as a normal power supply from the front panel with the MODE switch in the SUPPLY position. The amplitude of the added AC component is controlled by adjusting Ry. The value of this resistor should be greater than the minimum load resistance which the external signal source is suited to drive, since the impendance looking into terminals C.S. and A4 is virtually zero ohms.

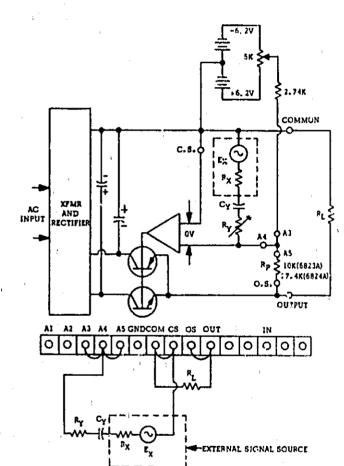


Figure 3-12. Constant Voltage Supply With Superimposed AC Output

3-53 The amplitude of the ac Output component is given by  $E_{AC} = E_X (R_p/R_X + R_y)$ , where  $R_X$  is the internal (Thevenin equivalent) source resistance. The capacitor Cy is added in order to block the DC current path through the oscillator, and must

be selected large enough to pass the lowest frequency of interest. The low frequency 3 db down point will be at  $_f = \frac{1}{2\pi(R_X + R_Y) \ C_Y}$ 

3-54 CONSTANT VOLTAGE AMPLIFIER WITH RE-MOTE GAIN CONTROL

3-55 The PS/A can be used as a power amplifier with Constant Voltage output that can be remotely controlled in a linear manner if the strapping arrangement shown on Figure 3-13 is employed. With this configuration, the gain is a linear function of control rotation, and the gain control is a linear function of control rotation, and the gain control is external to the PS/A. Ex and Rx are the equivalent (open circuit) voltage and output resistance looking back into the external signal source. Ry must be selected to have a value at least as large as the minimum load resistance which the external signal source can feed without overloading or distorting. Cy is inserted in order to block any small DC currents which otherwise might flow from the oscillator, producing a DC component on the output. Cy must be selected large enough to pass the lowest frequency of interest; the 3 db down frequency is

$$f = \frac{1}{2\pi (R_X + R_Y) C_Y}$$

3-56 If it is desired to amplify a DC signal, or one which has both a DC and AC component, then the capacitor  $C_Y$  must be eliminated from the circuit.

3-57 The output voltage is given by the equation

$$E_O = E_X \frac{Rp}{R_X + R_Y}$$

Proper impedance levels will be maintained if approximately 2 mA is allowed to flow through Ry and Rp. Thus wherever possible, Ry should be selected so that  $R_{Y} = \frac{E_{X}}{2(10^{-3})} - R_{X}$ 

where the Ex for this equation is the maximum <u>peak</u> <u>value</u> which will be amplified. This equation, of course, assumes that the external signal source is capable of delivering 2 mA without overloading or distorting, and should be altered if less current is available.

3-58 <u>Current Input</u>. Actually, the components  $E_X$ ,  $R_X$ , and  $R_Y$  of Figure 3-13 comprise a current source feeding the summing point A4 and the gain control  $R_p$ . A current source can be substituted in place of these components, and the output voltage

will be V<sub>OUT</sub> = IXRP, where IX is the current source connected between CS and A4. As shown in Pigure 3-13. The capacitor Cy should also be added in series if it is desired to suppress any DC component present in the input signal source.

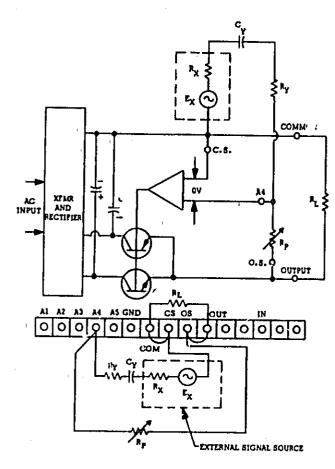


Figure 3-13. Constant Voltage Amplifier With Linear Remote Gain Control

3-59 Two or More Inputs. The concept of Figure 3-13 is readily adapted, using standard operational amplifier techniques, to the simultaneous amplification of two or more signals. Figure 3-14 shows the nanner in which two voltage inputs can be mixed and amplified; additional inputs can be added in parallel without limit, and current input signals can be substituted for voltage input signals, as discussed in the previous paragraph. The capacitors Cy and C'y are only included when it is desired to:

- a. Prevent the DC input level of the signal from affecting the DC level of the output, or
- b. Prevent the DC level of the PS/A from being fed back into the output terminals of the signal source.

3-60 The output for two (or more) inputs is given by:  $E_O = E_X \frac{RP}{RX + RY} + E_X' \frac{RP}{RX' + RY'} + \dots$ 

Thus Ry, Ry' can be made variable and used as relative level controls while Rp functions as a master gain control. Notice that any of the signal inputs which are to be mixed and amplified may be ac, dc, or both.

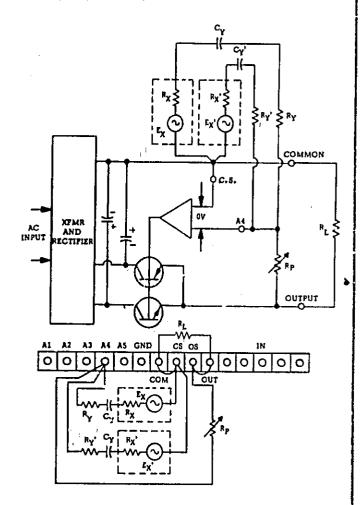


Figure 3-14. Constant Voltage Amplifier, Two Inputs

# 3-61 CONSTANT CURRENT OPERATING MODES

3-62 In order to achieve constant current operation, it is necessary to utilize an external current sampling resistor which yields a voltage drop that is proportional to the output current. This voltage drop can then be applied to the comparison amplifier in the input circuit and compared with a fraction of the reference voltage (or input voltage). The output of the comparison amplifier is then fed back to the regulator/output amplifier in a manner identical to that occurring during constant voltage operation. Hence, the only conceptual difference

between Constant Current operation and Constant Voltage operation is that the voltage which is sensed by the compactson amplifier is taken across the current sampling resistor rather than across the load.

3-63 The regulatior ripple, temperature coefficient, and stability of Constant Current Power Supply/Amplifier performance can be predicted by dividing the corresponding Constant Voltage specification by the ohmic value of  $R_{\rm S}$ . For example, the use of a 2 ohm current sampling resistor converts the 2 mV rms Constant Voltage ripple and noise specification of the 6823A to 1 MA, etc. In the case of temperature coefficient and stability, it is also necessary to add in the percentage effect due to any change in ohmic value of the monitoring resistor  $R_{\rm S}$ .

3-64 The output impedance of the Constant Current source at DC is given by the relationship  $Z_O = (\Delta E_L/\Delta I_L)$ , where  $\Delta E_L$  is the change in load voltage associated with a load resistnace change, and  $\Delta I_L$  is the small resulting output current change. The fact that the PS/A has a high loop gain and no output capacitor means that the PS/A used in a Constant Current mode will have a high output impedance. In fact, it can be shown that the impedance at any frequency for a Constant Current source of this type will be approximately equal to  $R_{\mathbf{M}}$  (AB), where AB is the loop gain of the feedback amplifier within the power supply at the frequency of interest. Since the PS/A has a bandwidth in excess of 20 KHz, this output impedance will remain quite high over a wide frequency band, Stated in terms of the time domain, the PS/A will respond quickly to changes in load resistance the load current transient will be short,

# 3-65 SELECTING RS

3-66 Particular care must be paid to the manner in which the current sampling resistor, Rg, is selected and employed. Constant Current operation, on a percentage basis, can be no better than this resistor.

3-67 R<sub>S</sub> is selected to yield a 1 volt drop at the maximum current rating of the instruments; its

maximum dissipation is 1/2 watt for the Model 6823A and 1 watt for the Model 6824A. It is therefore recommended that a minimum 30 watt wirewound resistor be used for the former and 60 watt for the latter. This ensures that the surface temerature of the resistor will not be high compared to the ambient, and therefore subject to long-term warm-up effects and short-term variations resulting in output "bobble". In any case, the temperature coefficient of the resistor should be 20 ppm/°C or less.

3-68 If it is intended to use the PS/A at less than maximum rated output curre .., proportionately higher values of Rs may be chosen. This will result in better performance at low currents. It should be remembered, however, that the resistor Rs cannot be increased in ohmic value indiscriminately for two reasons:

- a. It soon becomes impractical to increase the wattage rating of RS by the necessary amount.
- b. The IR drop which occurs across Rs detracts! directly from the power supply voltage rating available to the load device.

3-69 Resistor Rs should be a four terminal device connected as shown on Figure 3-15. Note that for convenience and clarity, the remaining Constant Current diagrams do not show this connection pattern explicitly.

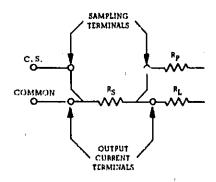


Figure 3-15. Current Sampling Resistor,
Connection Method

# 3-70 NORMAL OPERATING MODES

3-71 Figure 3-16 shows the connection pattern and a sociated circuit configuration for operating the S/A as a continuously variable, reversible P. Constant Current source. This is the Constant Current counterpart of the Constant Voltage mode of operation shown in Figure 3-2. The MODE switch is in the SUPPLY position.

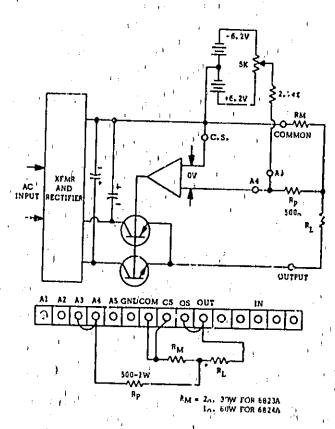


Figure 3-16. Standard Constant Current Power Supply

3-72 Figure 3-17 shows the simplest way of operating the PS/A as a Constant Current output amplifier with continuously variable gain. The front panel FINE/ZERO control should be adjusted for zero do output current (assuming the input source has no do input component). With the values given in Figure 3-17, the external oscillator must have a voltage output of 1.5 volts peak in order to drive the power amplifier to its full output current car billity. The impedance saen by this external voltage source will be approximately 1.7K.

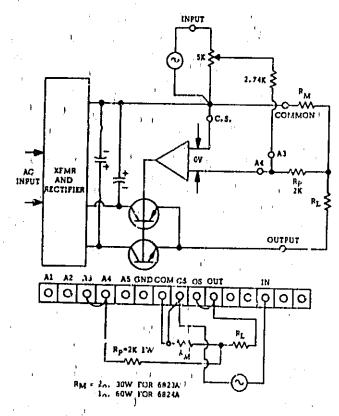


Figure 3-17. Standard Constant Current Amplifier

# 3-73 REMOTE PROGRAMMING, BIPOLAR OUTPUT

3-74 Figure 3-18 shows one method of controlling the output of the PS/A as a remote programmed Constant Current power supply. Switch S1 causes the output current to be either positive or negative. If it is desired to achieve only a positive or only a negative output current (unipolar output), then \$1 must be eliminated. Terminal A4 is strapped directly to A2 for a positive output, or directly to A1 for a negative output.

3-75 The front panel PINE/ZERO control permits the output current to be adjusted to zero when Rp is set to zero, while the slope of the programming coefficient is adjusted by means of R38 (positive output) and R39 (negative output). Either of the switched resistance schemes of Figure 3-10 may be substituted for a variable Rp in this or any other Constant Current diagrams which follow. The advantage of the circuit of Figure 3-18 is that the output current is a linear function of the programming resistance; the disadvantage is that a switch must be used in order to accomplish polarity reversing.

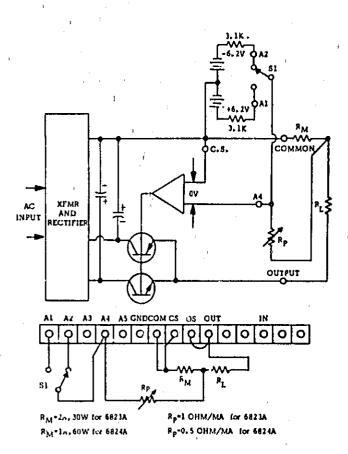


Figure 3-18. Remote Resistance Programming, Bipolar Output (CC)

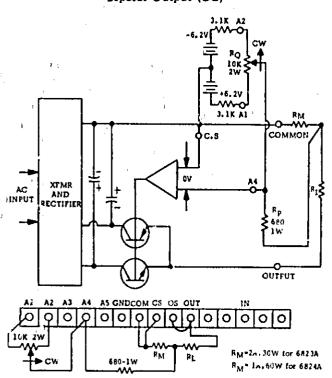


Figure 3-19. Switchless Bipolar Programming (CC)

# 3-76 SWITCHLESS BIPOLAR PROGRAMMING

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3-77 Figure 3-19 shows the method of connecting the PS/A for Constant Current bipolar programming without the necessity for polarity switching. The 680 ohm resistor should be wire-wound with a temperature coefficient of less than 20 ppm/°C. Switched resistances can be substituted for resistor RQ in Figure 3-19. The disadvantage of this mode lies in the non-linear (but accurately reproducible) programming characteristic.

# 3-78 CONSTANT CURRENT SUPPLY WITH SUPER-IMPOSED AC OUTPUT

3-75 Figure 3-20 shows the method of adding an AC component on top of the adjustable DC output current. The dc level is controlled in the normal fashion from the front panel. The amplitude of the added ac component is determined by the value of Ry, which, however, should not have a value less than the minimum load resistance which the external signal source is capable of driving without overloading or distorting.

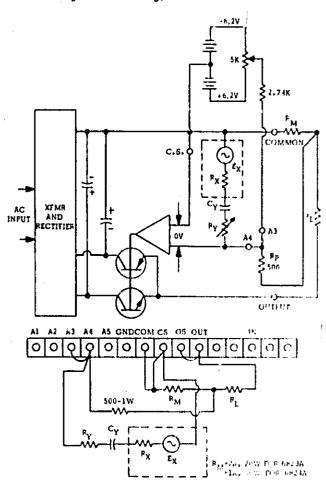


Figure 3-20. CC Power Supply With Superimposed AC Output

3-80 The peak voltage Ex which is necessary to achieve a peak output current equal to the maximum rating of the instrument, equals  $(R_X + R_Y) / R_P$ . The capacitor  $C_Y$  is added to block the DC current path through the oscillator.  $C_Y$  must be selected large enough to pass the lowest frequency of interest; the low frequency 3 db down point will be at

$$f = \frac{1}{2\pi (R_X + R_Y) S_Y}$$

where Rx is the internal (Thevenin equivalent) source resistance. Of course, capacitor  $C_Y$  is not employed if it is desired for the Constant Current output to respond to the dc component of the external signal source.

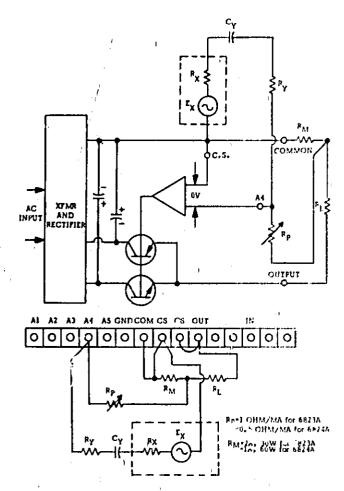


Figure 3-21. CC Amplifier With Linear Remote Gain Control

3-81 CONSTANT GURRENT AMPLIFIER WITH REMOTE GAIN CONTROL

3-82 Figure 3-21 shows the method of connecting an external voltage source with a PS/A to obtain Constant Current output with linear remote gain control. Using the recommended values for RS, the output current is given by:

$$I_O = (E_X/R_M) \cdot R_P/(R_X + R_V).$$

It is recommended that Ry be selected so that at the peak input voltage  $E_X$  the current flowing through  $R_X$ ,  $R_Y$ , and  $R_P$  is 2 MA. With this choice,  $R_P = 500$  ohms results in an output current equal to the maximum rating of the instrument.

3-83 Cy is chosen large enough to pass the lowest frequency of interest and is omitted if it is desired to have the output current responsive to the DC (as well as AC) component of the input voltage Ex.

3-84 Care should be taken that Ry is chosen to be more than the minimum load resistance which the external signal source can feed without overload or distortion. However, Ry should not be chosen too large, or the current flowing through Rp will be influenced by the current into the composition amplifier. Thus it is undesirable to select Ry so that the current flowing through it with full input signal is less than 1 MA.

3-85 Current Input. The combination of components Ex, Rx, and Ry can always be replaced by a Constant Current source. If this is done, the current gain of the configuration of Figure 3-21 is:

$$\frac{IOUT}{I_{IN}} = \frac{Rp.}{R_{M}}$$

3-86 Two or More Inputs. If desired, the circuit of Figure 3-22 can be used to add the effect of two or more signals. All the equations given in this section hold true for each of the multiple inputs, with no interaction.

$$IO = \frac{E_X}{R_M} - \frac{R_P}{R_X + R_Y} + \frac{E_{X'}}{R_M} - \frac{R_P}{R_{X'} + R_{Y'}} + \dots$$

# 3-87 SERIES AND PARALLEL OPERATING MODES

3-88 The following paragraphs include methods for combining separate PS/A's in series and parallel combinations. These methods are employed whenever it is required to extend the voltage or current capability beyond that available from one instrument alone. In all of the following diagrams the strapping pattern and circuit configuration for slave units have been shown in complete detail, In several of the diagrams, however, the strapping pattern and circuit configuration have not been completed for the Master unit. This was done because the configuration chosen for the Master is determined by the manner in which it is desired for the ensemble of units to operate, and may be selected from any of the diagrams already given in this Section. Whether a combination acts as a power supply or amplifier, Constant Voltage or Constant Current source, locally or remotely controlled by resistance, voltage, or current, depends entirely on the connection mode selected for the Master Unit.

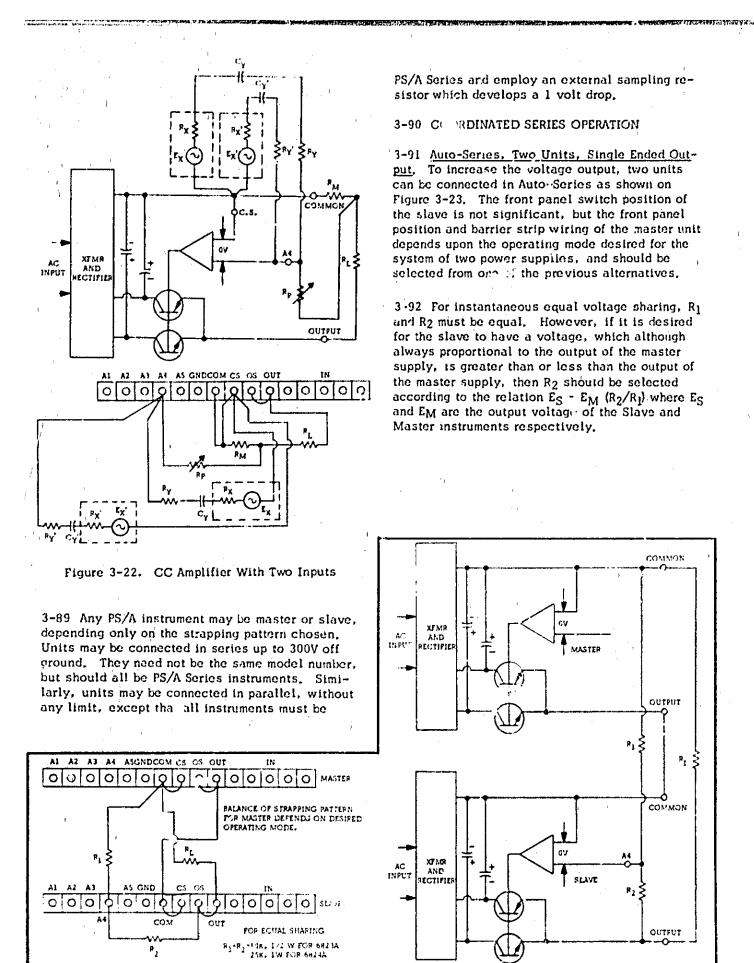


Figure 3-23. Auto-Series Operation, Two Units

3-93 Auto-Series, Three or More Units. The concept of the previous paragraphs is easily extended to three supplies as shown in Figure 3-24. If un-

COMMON XFMR AND RECTIFIES AC INPUT MASTER OUTPUT COMMON XFMR AND RECTIFIE AC. SLAVE #1 R<sub>2</sub> OUTPUT COMMON XFMR AND INPILE RECTIFIER SLAVE #2 OUTPUT AS GNDCOM CB OS MASTER 00000000000 000 BALANCE OF STRAPPING PAT. TRN FOR MASTER DEPENDS ON DESIRED OPERATING MODE. 0 STAVE #2 00 0 R<sub>1</sub>-R<sub>2</sub>-R<sub>3</sub>=10K, 1/2 W FOR 5823A 25K, 1W FOR 6824A

Figure 3-24. Auto-Series Operation, Two Units

equal voltage contributions are desired, the output of Slave #1 is given by  $E_{S1} = E_{M} (R_2/R_1)$ ; for Slave #2,  $E_{S2} = E_{14} (R_3/R_1)$ .

3-94 The concept of Figure 3-24 is easily extended to series combinations involving more than three units by merely iterating that portion of the circuit configuration which is bounded by the two horizontal dotted lines.

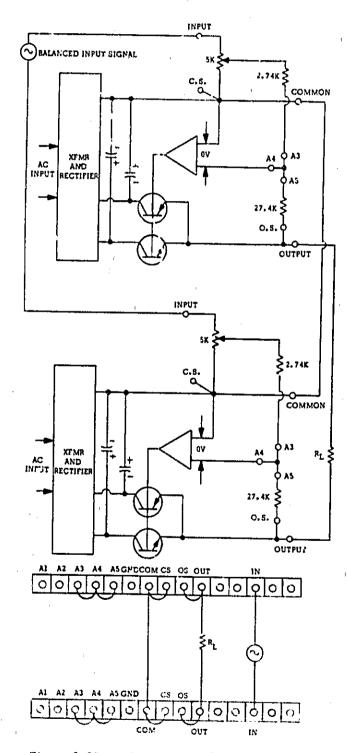
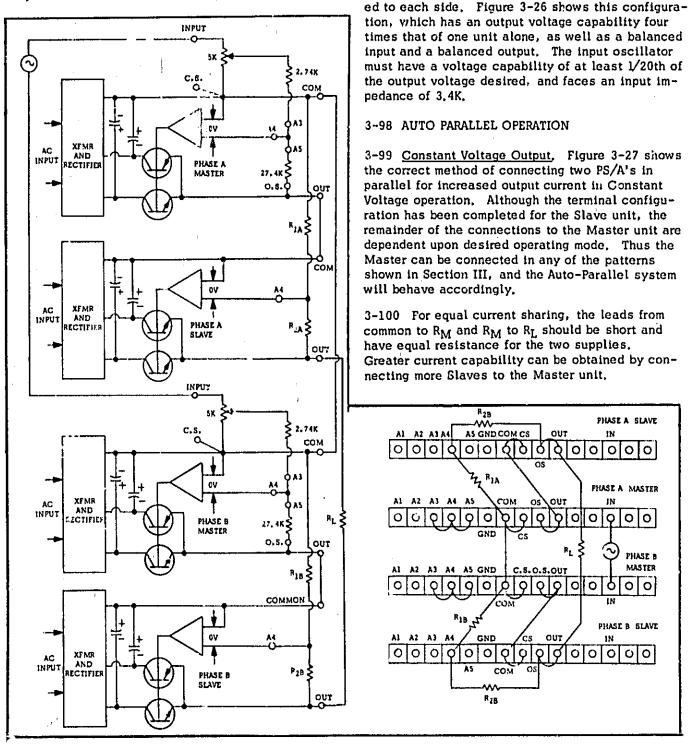


Figure 3-25. Balanced Input/Output Series Operation, Two Units

3-95 Series Operation, Balanced Output. The previous Auto-Series configurations are accomplished with single-ended input and single ended output having a common terminal. In some applications, however, it is necessary to employ an amplifier with input and output signals balanced about ground. Figure 3-25 shows the circuit configuration and strapping pattern for this mode of operation.



3-96 Since this mode is more likely to be employed as an amplifier rather than as a power supply,

the diagram shows an oscillator input. The voltage input to this circuit must be at least V10th

3-97 The balanced input concept can be extended

to even higher voltages if voltage Slaves are add-

the desired voltage output; the input impedance

seen by the oscillator is approximately 3.4K.

Figure 3-26. Balanced Input/Output Series Operation, Four Units

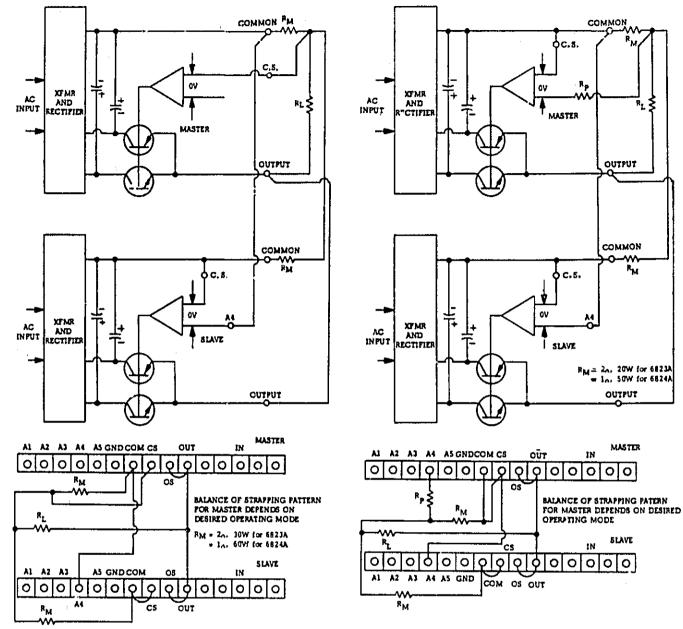


Figure 3-27. Auto-Parallel Operation, Constant Voltage Output

3-101 Constant Current Output. Figure 3-28 shows the correct method of connecting two supplies in Auto-Parallel for Constant Current cutput. The strapping configuration and circuit diagram for the master unit are not complete but can be wired

Figure 3-28. Auto-Parallel Operation, Constant Current Output

in accordance with any of the diagrams given in Section III and the Auto-Parallel combination will behave accordingly. Still greater current output can be accomplished by adding any desired number of slaves to the master in the same way.

# SECTION IV PRINCIPLES OF OPERATION

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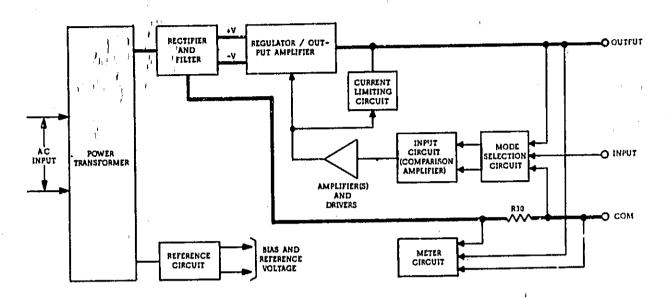


Figure 4-1. Overall Block Diagram

# 4-1 OVERALL BLOCK DIAGRAM DISCUSSION (Figure 4-1)

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**3** iii

4-2 The Power Supply/Amplifier (PS/A) can be operated as either a power supply or a power amplifier. The operating mode is determined by the mode selection circuit which couples to the input circuit; either the voltage across the output terminals (during power supply operation) or the voltage across the input terminals (during amplifier operation).

#### 4-3 POWER SUPPLY OPERATION

- 4-4 During power supply operation, the PS/A functions as a well-regulated dc power supply capable of furnishing a bi-polar. Constant Voltage output. Constant Current operation can also be achieved by connecting an external current sampling resistor of the unit.
- 4-5 The ac line voltage is reduced to the proper level by the power transformer and coupled to the rectifier and filter. The rectifier-filter converts the ac input to raw dc which is fed to the regulator/output amplifier circuit in both positive and negative form. The regulator/output amplifier acts as a series regulator during power supply operation. Its conduction is altered in accordance with the feedback control signals received from the driver

amplifiers, thus maintaining the output voltage (or current) constant. The feedback control signals also determine the polarity of the output voltage. For positive power supply output voltages the regulator circuit receives a positive input and for negative output voltages it receives a negative input. Hence, the dc output voltage can be continuously adjusted from its maximum rated positive value to an equal negative value.

- 4-6 The current limiting circuit monitors the output current passing through the regulator. If this current exceeds a certain preset limit, the limiting circuit conducts, shunting input current away from the regulator and thus keeping the output current constant.
- 4-7 The input circuit serves as the first "link" in the feedback circuit. It detects any changes in the output voltage and sends a feedback control signal to the series regulator via the amplifierdrivers. The feedback signal is of the correct phase and amplitude to counteract the change in output voltage. Notice that the supply can also be used as a Constant Current source as outlined in Section III of this manual. Under these conditions, the input circuit monitors the voltage dropped across an externally connected current sampling resistor and, in this manner develops the feedback error signals necessary to maintain the output current constant.

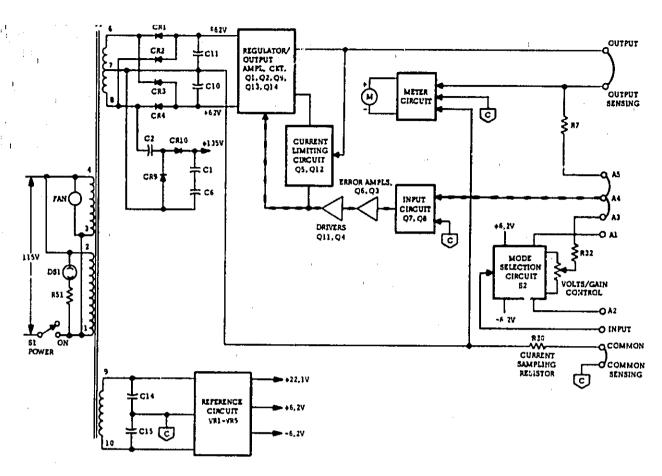


Figure 4-2, Simplified Schematic

4-8. The reference circuit provides stable reference voltages which are used throughout the unit for biasing and comparison purposes. The meter circuit provides indications of dc output voltage or current. When measuring voltage it is connected across the output of the supply; when measuring current it is connected across an internal current sampling resistor, R30.

# 4-9 AMPLIFILR OPERATION

- 4-10 During amplifier operation, the unit acts as a power amplifier with a single-ended, push-pull output stage. It furnishes either a Constant Voltage output or a Constant Current output (with the addition of an external current sampling resistor). It is capable of amplifying both ac and dc input signals over a bandwidth of dc to 20 kHz.
- 4-11 Input signals are received from the external source via the input and common terminals. The input circuit and amplifier-drivers amplify the signal and forward it to the regulator/output amplifier circuit. This circuit functions as a push-pull output amplifier during amplifier operation, although note that no internal connections are necessary in the regulator/output amplifier circuit in order to accomplish this.

- 4-12 The current limiting circuit is operable in the amplifier mode and serves to protect both the amplifier and the load by preventing the output current from exceeding the preset current limit.
- 4-13 In a manner similar to that used during power supply operation, the input circuit generates the error signals necessary to keep the output voltage (or current) constant despite variations in the load or line circuits. The meter circuit for Model 6824A units is capable of indicating ac output voltage during amplifier operation, while Model 6823A meter circuits can measure do output voltage only.

# 4-14 SIMPLIFIED SCHEMATIC

4-15 A simplified schematic of the PS/A is shown in Figure 4-2. It shows the internal sources of bias and reference voltages and their nominal magnitudes with an input of 115Vac and under full load. It also shows some of the operating controls; the off-ON pushbutton (S1) and the VOLTS/GAIN control (R52A). METER switch S3, included in the meter circuit block on Figure 4-2, permits the meter to read dc output voltage or current, or ac output voltage. Rectifier diods CR1 through CR4, and filter capacitors C10 and C11 provide

+62Vdc and -62Vdc to the regulator/output amplifier circuit. These voltages are also fed to the error amplifiers and drivers for biasing purposes. Another bias voltage (+140V) is obtained from a voltage doubler network composed of CR9, CR10, C1, C2, and C6.

4-16 MODE switch S2, in the mode selection circuit, connects the VOLTS/GAIN potentiometer between a bi-polar reference source (±6.2V and -6.2V) during power supply operation, or across the external input signal path during amplifier operation. Hence, the position of the VOLTS/GAIN potentiometer determines the polarity and magnitude of the output voltage in the power supply mode or the gain of the PS/A in the amplifier mode.

4-17 Notic that the PS/A does not have the large output capacitor connected across most power supplies to insure feedback loop stability. This capacitor has been removed to ease the bandwidth limitations during amplifier operation. Removal of the output capacitor has the additional advantage of greatly enhancing the programming speed during power supply operation. Alternate methods of insuring feedback loop stability are employed throughout the PS/A feedback circuit to offset the absence of this capacitor.

# 4-18 <u>DETAILED CIRCUIT ANALYSIS (See schematic</u> at rear)

# 4-19 INPUT CIRCUIT

4-20 The input circuit consists basically of a differential amplifier stage (Q7 and Q8), and the FINE/ZERO and VOLTS/GAIN controls (R52A and R52B). Note that for simplicity's sake, R52A has been included in the mode selection circuit on the schematic.

4-21 The input circuit controls the conduction of the series regulator/output amplifier transistors. To accomplish this, the circuit continuously compares a fraction of 'he output voltage (or current) with a fixed reference voltage and, if a difference exists, produces an "error" voltage whose amplitude and phase is proportional to the difference. The "error" output is fed back to the regulator/output amplifier, via the error and driver amplifiers. The feedback voltage alters the conduction of the regulator/output amplifier transistors which. in turn, alter the output voltage (or current) so that the difference between the two differential amplifier input voltages is reduced to zero. The above action not only maintains a constant output regardless of line or load variations; but establishes output levels in accordance with the setting of the VOLTS/GAIN control.

4-22 Stage Q7 of the differential amplifier is connected to a voltage divider consisting of R40, R52B, R47, and R48. The FINE/ZERO control, R52B, permits the bias at the base of Q7 to be adjusted slightly above or below the common sensing (C) point. During power supply operation the FINE/ZERO control provides a fine adjustment of the dc output voltage or current and during amplifier operation it provides a dc "offset" adjustment of approximately 1% of the maximum rated output. Hence, the average value of an ac output signal can be adjusted exactly to zero with this control.

4-23 Stage Q8 is connected to the voltage (or current) summing point (A4) at the junction of the VOLTS/GAIN control and resistor R7. The potential at the summing point will instantaneously change if the output voltage/current attempts to change or if the voltage that is picked off by R52A changes. Output variations affect the voltage dropped across R7 thus changing the summing point potential. Similarly, moving the VOLTS/ GAIN control arm affects the summing point potential. During power supply operation, this control is connected across a bi-polar reference source. Moving the control toward the positive end of the potentiometer causes the summing point to go positive resulting in a more negative output. Conversely, moving the control toward the negative end results in a more positive output. During amplifier operation the external input signal is applied across R52A and moving the control arm varies the amount of signal that is picked off,

4-24 Summing point variations are felt at the base of the Q8 which varies its conduction in accordance with the polarity and magnitude of the change at the summing point. The resulting error voltage is taken from the collector of Q8 and ultimately alters the conduction of the regulator/amplifier transistors.

4-25 Feedback network C8, R25 helps stabilize the feedback loop. Diodes CR25 through CR28 form a limiting network which prevent excessive voltage excursions from over-driving stage Q8. Resistor R45 in series with the base of Q8 is a current limiting resistor.

# 4-26 MODE SELECTION CIRCUIT

4-27 The mode selection circuit consists basically of MODE switch S2 which determines the operating mode of the instrument. As mentioned previously, S2 connects the VOLTS/GAIN control between a bipolar reference source in the SUPPLY position and across the input signal path in the AMPLIFIER position. In the AMPLIFIER position S2 also connects a load resistor (R57) between the bi-polar reference voltages to prevent loading down the reference circuit.

# 4-28 ERROR AND DRIVER AMPLIFIERS

4-29 The error and driver amplifies amplify the error signal from the input circuit to a level sufficient to drive the regulator/amplifier transistors. Amplifier stages Q6 and Q3 each contain a stabilization network (C5-R24 and C7-R14, respectively) which provides for roll off in the loop gain. Diode CR11 clamps the emitter of Q6 to a stable reference voltage, +22.1V. Emitter follower Q4 drives the regulator/amplifier transistors Q1, Q2, Q9, Q13, and Q14.

# 4-30 REGULATOR/OUTPUT AMPLIFIER CIRCUIT

4-31 During power supply operation, transistors Q1, Q9 and Q2, Q13, Q14 serve as series control elements in the positive and negative output lines, respectively. The conduction of the series transistors is controlled by the feedback voltage from Q4. Note that when the feedback voltage is negative, PNP transistors Q2,Q13 and Q14 are conducting and NPN transistors Q1 and Q9 are cut-off. Under these conditions, the supply furnishes a negative output voltage. The reverse is true when the feedback voltage is of positive polarity. Transistor Q10 furnishes a constant bias current to the bases of the series transistors. At zero volts output, both banks of series transistors are conducting slightly because of the base current provided by Q10. Diodes CR18 and CR19 provide the voltage drop necessary to forward bias each regulator bank simultaneously. This eliminates any "dead spots" when the output voltage is programmed away from zero. Note, however, that as soon as the output voltage is driven away from zero, one bank of transistors remains in conduction while the other is cutoff.

4-32 During amplifier operation, the transistors serve as a single-ended, push-pull, output amplifier. Although the schematic at the rear of the manual shows Q1, Q2, Q9, Q13, and Q14 drawn as a conventional series regulator, a closer inspection will reveal that the circuit could be redrawn as a push-pull amplifier without changing any of the connections. The output amplifiers are biased for class AB operation and connected in a complementary configuration, as described in the previous paragraph, thus obviating the need for center-tapped input and output transformers.

# 4-33 CURRENT LIMITING CIRCUIT

4-34 The current limiting circuit consists of transistors Q5 and Q12 and associated resistors.

These transistors keep the output current from exceeding the current limit established by the setting of potentiometers R5 and R6. PNP transistor O5 limits the current in the negative line and NP<sub>1N</sub> transistor Q12 limits the current in the positive line. Under increased current demands, the current through the PNP regulator/amplifier transistors increases, increasing the voltage drop across R6 and driving the base of Q5 in a negative direction. The conduction of Q5 limits the base current of the PNP transistors and, hence, the output current. Operation of current limiting transistor Q12 is similar; except that the base current of the NPN transistors is limited further back in the feedback loop at the input of Q4.

# 4-35 REFERENCE CIRCUIT

4-36 The reference circuit is an auxiliary supply which provides stable reference voltages used throughout the instrument for biasing and comparison purposes. The reference voltages are developed across temperature compensated Zener dides VR1 through VR5. The reference voltages are all derived from smoothed positive and negative do obtained from full wave rectifiers CR5, CR6 and CR7, CR8 and filter capacitors C12 and C13.

### 4-37 METER CIRCUIT

4-38 The meter circuit provides continuous indications of output voltage or current on a single meter. The meter can be used as a dc voltmeter or ammeter or as an ac voltmeter depending upon the position of METER switch S3. With S3 in the dc VOLTS position the meter is connected in series with R31 and R33 across the output of the supply. Resistor R31 permits dc voltage calibration of the meter to compensate for slight resistance variations.

4-39 With S3 in the DC AMPS position, the meter is connected across current sampling resistor R30, whose voltage drops varies in proportion to the output current. Current calibrate potentiometer, R35 is adjusted for full scale deflection in the current range.

4-40 In the AC VOLTS position, the meter indicates the rms output voltage of the amplifier. Rectification of the ac output voltage is accomplished by diodes CR13 through CR16.

4-41 The meter movement used in this instrument, can withstand a current overland of many times the maximum rated output without damage.

# MAINTENANCE

# SECTION V MAINTENANCE

# 5-1 INTRODUCTION

Upon receipt of this instrument, the performance check (Paragraph 5-10) should be made. This check is suitable for incoming inspection. If a fault is detected in the unit while making the performance check or during normal operation. proceed to the troubleshooting procedures (Paragraph 5-24). After troubleshooting and repair (Paragraph 5-34), perform any necessary adjustments and calibrations (Paragraph 5-36). Before returning the PS/A to normal operation, repeat the performance check to ensure that the fau. r has been properly corrected and that no other faults exist. Before doing any maintenance checks, turn-on power supply, allow a half-hour warm-up, and read the general information regarding measurement techniques (Paragraph 5-3).

# 5-3 GENERAL MEASUREMENT TECHNIQUES

- 5-4 The measuring device must be connected across the sensing leads of the unit or as close to the output terminals as possible when measuring the output impedance, transient response, regulation, or ripple of the power supply in order to achieve valid measurements. A measurement made across the load includes the impedance of the leads to the load and such lead lengths can easily have an impedance several orders of magnitude greater than the supply impedance, thus invalidating the measurement.
- 5-5 The monitoring device should be connected to the OS and OS terminals or as shown in Figure 5-1. The performance characteristics should never be measured at the front terminals if the load is connected across the rear terminals. Note that when measurements are made at the front terminals, the monitoring leads are connected at A, not B; as shown in Figure 5-1. Failure to

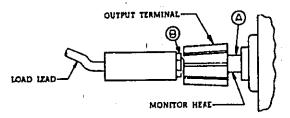


Figure 5-1. Front Panel Terminal Connections

connect the measuring device at A will result in a measurement that includes the resistance of the leads between the output terminals and the point of connection.

- 5-6 For output current measurements, the current sampling resistor should be a four-terminal resistor. The four terminals are connected as shown in Figure 3-15. In addition, the resistor should be of the low noise, low temperature coefficient (20ppm/OC or less) type and should be used at no more than 5% of its rated power so that its temperature rise will be minimized.
- 5-7 When using an oscilloscope, ground the common terminal of the PS/A and then ground the case of the oscilloscope to this same point. Make certain that the case is not also grounded by some other means (such as the power line). Connect both oscilloscope input leads to the PS/A ground terminal and check that the oscilloscope is not exhibiting a ripple or transient due to ground loops, pick-up, or other means.

# 5-8 TEST EQUIPMENT REQUIRED

5-9 Table 5-1 lists the test equipment required to perform the various procedures described in this Section.

Table 5-1. Test Equipment Required

Туре	Required Characteristics	Use	Recommended Model
Differential Voltmeter	Sensitivity: 1 mv full scale (min.), Input impedance: 10 megohms (min.),	Measure DC voltages; calibration procedures	₹ 3420 (See Note)
Variable Voltage Transformer	Range: 90-130 volts, Equipped with voltmeter ac- curate within 1 volt,	Vary AC input	
AC Voltmeter	Accuracy: 2%, Sensitivity: 1 mv full scale deflection (min.).	Measure AC voltages and ripple.	<i>№</i> 403 B
Oscilloscope	Sensitivity: 10 μv/cm. Dif- ferential input.	Peak-to-peak measurements. Waveform displays.	f 140 A plus 1400 A plug in.
Oscillator	Range: 5 Hz to 600KHz Accuracy: 2%	Impedance Checks, amplifier input source	№ 200 CD
Square Wave Generator	Rise time: 0,02 μsec Frequency: 1 Hz to 1 MHz	Measure programming speed	₩ 211 A
Repetitive Load Switch	Rate: 60-400 Hz, 2µsec rise and fall time.	Measure transient asponse	See Figure 5-5
Resistor	Value: See Paragraph 5-13 and Figure 5-3, ±5%, 75 watts	Load Resistor	
Resistor	Value: See Figure 5-3. 1%, 30 watts (6823A) 60 watts (6824A) 20ppm, 4-Terminal.	Current sampling	
Resistor	1 Ka ±1%, 2 watt non-inductive	Measure impedance	*
Resistor	100 ohms, ±5%, 10 watt	Measure impedance	
Capacitor	500μf, 50 wvdc	Measure impedance	

# NOTE

A satisfactory substitute for a differential voltmeter is to arrange a reference voltage source and null detector as shown in Figure 5-2. The reference voltage source is adjusted so that the voltage difference between the supply being measured and the reference voltage will have the required resolution for the measurement being made. The voltage difference will be a function of the null detector that is used. Examples of satisfactory null detectors are: If 419A null detector, a DC coupled oscilloscope utilizing differential input, or a 50 mv meter movement with a 100 division scale. For the latter, a 2 mv change in voltage will result in a meter deflection of four divisions.

# CAUTION

Care must be exercised when using an electronic null detector in which one input terminal is grounded to avoid ground loops and circulating currents.

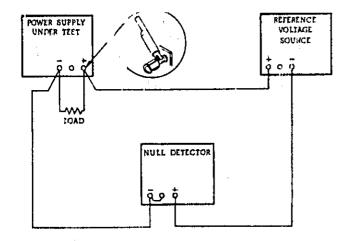


Figure 5-2. Differential Voltemter Substitute,
Test Setup

# 5-10 PERFORMANCE TEST

5-11 The following test can be used as an incoming inspection check and appropriate portions of the test can be repeated either to check the operation of the instrument after repairs or for periodic maintenance tests. The tests are performed using a 115-VAC 60 Hz, single phase input power source. They test the performance of the unit both as a power supply and as a power amplifier. If the correct result is not obtained for a particular check, do not adjust any controls; proceed to troubleshooting (Paragraph 5-24).

# 5-12 POWER SUPPLY TESTS

- 5-13 Rated Output and Meter Accuracy. To check the output voltage, proceed as follows:
- a. Connect 40 ohm load resistor (50 ohms for Model 6824A) across rear output terminals of supply.
- b. Connect differential voltmeter across OS and CS terminals of sumply observing correct polarity.
- c. Set METER switch to VOLTS and MODE switch to SUPPLY.
- d. Adjust VOLTS/GAIN control clockwise (cw) until front panel meter indicates exactly the maximum rated positive output voltage.
- e. Differential voltmeter should indicate maximum rated positive output voltage within ±2%.
- f. Adjust VOLTS/GAIN control in ccw direction until front panel meter reads maximum rated negative output voltage.

- g. Reverse position of input polarity switch on differential voltmeter. Voltmeter should indicate maximum rated negative output voltage within  $\pm 2\%$ .
- 5-14 To check the output current, proceed as follows:
  - a. Connect test setup shown in Figure 5-3.

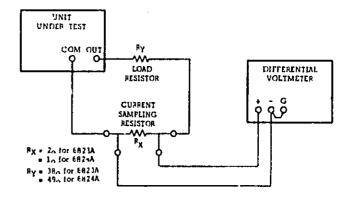


Figure 5-3. Output Current, Test Setup

- b. Set METER switch to AMPS.
- c. Turn on supply and adjust VOLTS/GAIN control cw until front panel meter indicates exactly maximum rated positive current.
- d. Differential voltmeter should read 1±0,02 Vdc.
- e. Reverse polarity switch on differential voltmeter and adjust VOLTS/GAIN control ccw for maximum rated negative current.
- f. Differential voltmeter should again read  $1\pm0.02$  Vdc.
- 5-15 Load Regulation, To check the Constant Voltage load regulation, proceed as follows:
- a. Connect test setup as shown in Figure 5-4.
  - b. Set METER switch to VOLTS,
- c. Turn on supply and adjust VOLTS/GAIN control cw until front panel meter indicates maximum rated positive output voltage.
- d. Read and record voltage indicated on differential voltmeter.
  - e. Disconnect load resistor,

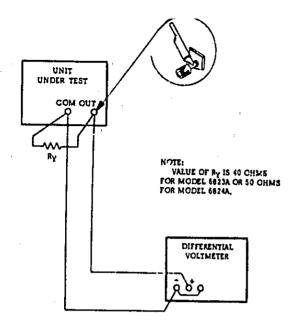


Figure 5-4. Load Regulation, Test Setup

- f. Reading on differential voltmeter should not vary from reading recorded in step d by more than 9 mVdc for Model 6823A or 15 mVdc for Model 6824A.
- g. Set input polarity switch to appropriate position. Rotate VOLTS/GAIN control ccw and repeat steps c through f for maximum rated negative output voltage.
- 5-16 <u>Line Regulation</u>. To check the Constant Voltage line regulation, proceed as follows:
- a. Connect variable auto transformer between input power source and PS/A power input.
  - p.') Connect test setup shown in Figure 5-4.
- c. Adjust variable auto transformer for 105 or 210 Vac input.
- d. Turn on supply and rotate VOLTS/GAIN control cw until front panel meter indicates exactly the maximum rated positive output voltage.
- e., Read and record voltage indicated on differential voltmeter.
- f. Adjust variable auto transformer for 125 or 250 Vac input.
- g. Reading on differential voltmeter should not vary from reading recorded in step e by more than 9 mvdc for Model 6823A units or 15 mvdc for Model 6824A units.

- h. Rotate VOLTS/GAIN control ccw to obtain maximum negative voltage and repeat steps d through g.
- 5-17 Ripple and Noise. To check ripple and noise, proceed as follows:
- a. Retain test setup used for previous line regulation test except connect AC voltmeter (\$\overline{\psi}\$ 403B) across output terminals instead of differential voltmeter.
- b. Adjust variable auto transformer for 125 Vac input.
- c. Turn on supply and adjust VOLTS/GAIN control cw until front panel meter indicates exactly the maximum rated positive output voltage.
- d. AC voltmeter should read less than 2 mV rms (Model 6823A) or 10 mV rms (Model 6824A).
- e. Reverse polarity of input to ac voltmeter and repeat steps c and d for maximum rated negative output voltage.
- 5-18 <u>Transient Recovery Time</u>. To check the transient recovery time of the supply, proceed as follows:
  - a. Connect test setup shown in Figure 5-5.

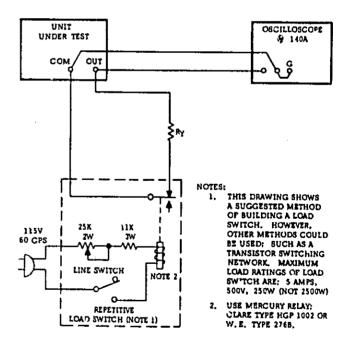


Figure 5-5. Transient Response, Test Setup

b. Rotate VOLTS/GAIN control cw to obtain maximum rated positive output.

- c. Close line switch on repetitive load switch setup.
- d. Adjust 25K potentiometer until a stable display is obtained on oscilloscope. Waveform should be within the tolerances shown in Figure 5-6 (output should return to within 5 mv of original value in less than 100 microseconds).

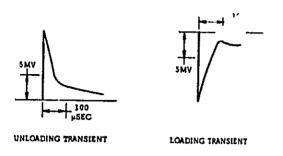


Figure 5-6. Transient Response Waveforms

- e. Reverse input connections to oscilloscope and repeat steps c and d for maximum rated negative output.
- 5-19 <u>Programming Speed</u>. To check the unit's programming speed, a square wave is applied to the unit and it is operated in the amplifier mode. This has the same effect as rapidly programming the unit, up and down, in the power supply mode. To make this test, proceed as follows:
  - a. Connect test setup shown in Figure 5-7.

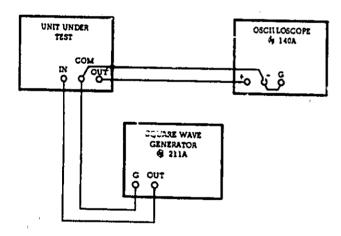


Figure 5-7. Programming Speed, Test Setup

- b. Set MODE switch to AMPLIFIER and turn on unit.
- c. Rotate VOLTS/GAIN control fully clock-wise.

- d. On pulse generator, set input frequency to about 2 kHz and adjust amplitude to obtain maximum rated peak-to-peak output signal on oscilloscope (-20V to +20V on Model 6823A and -50V to +50V on Model 6824A).
- e. Adjust oscilloscope to observe rise time of one square wave. The wave shape should be within the tolerances shown on Figure 5-8 (output should change from maximum rated negative value to maximum rated positive value in less than 50 µsec).

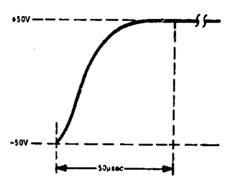


Figure 5-8. Typical Programming Speed Waveforms

- f. Check the fall time of one square wave. It should be almost identical to the rise time except for inversion.
- 5-20 Output Impedance. To check the output impedance, proceed as follows:
  - a. Connect test setup shown in Figure 5-9.

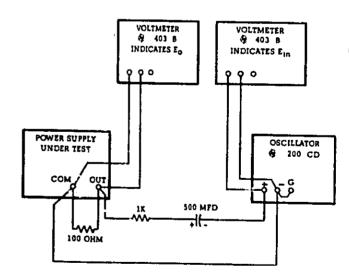


Figure 5-9. Cutput Impedance, Test Setup

b. Turn on supply and adjust VOLTS/GAIN control until front panel meter reads +10 volts (+20 volts for Model 6824A).

- c. Set AMPLITUDE control on Oscillator to 10 volts  $(E_{in})$ , and FREQUENCY control to 100Hz.
- d. Record voltage across output terminals of the power supply ( $E_{\rm O}$ ) as indicated on AC voltmeter.
- e. Calculate the output impedance by the following formula:

$$Z_{out} = \frac{E_{oR}}{E_{in} - E_{o}}$$

 $E_0 = rms$  voltage across power supply output terminals.

R = 1000

 $E_{in} = 10 \text{ volts}$ 

- f. The output impedance (Zout) should be less than 0.03 ohm.
- g. Using formula of step e calculate output impedance at frequencies of 500Hz and 100KHz. Values should be less than 0.3 ohm and 3.0 ohms, respectively.

### 5-21 AMPLIFIER TESTS

- 5-22 Gain and Frequency Response. To check the gain and frequency response of the amplifier, proceed as follows:
  - a. Connect test setup shown in Figure 5-10.

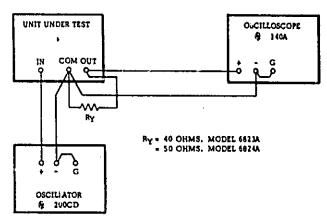


Figure 5-10. Gain and Frequency Response, Test Setup

- b. Set MODE switch to AMPLIFIER.
- c. Set oscillator frequency at 2KHz and output at 4 Vac, peak-to-peak (Model 6823A) or 10 Vac, peak-to-peak (Model 6824A).
- d. Adjust VOLTS/GAIN control on amplifier to obtain 40 volt peak-to-peak (Model 6823A) or 100 volt peak-to-peak (Model 6824A) reading on

oscilloscope. Reading should be within  $\pm 2\%$  tolerance.

- e. Adjust oscillator frequency to 20 KHz while maintaining same oscillator output voltage.
- f. Waveform on oscilloscope should not be less than 29 volts peak-to-peak (Model 6823A) or 71 volts peak-to-peak (Model 6824A).
- 5-23 AC Meter Accuracy (Model 6824A Only). The front panel meter on Model 6824A units is capable 1 measuring ac output voltages. To check its accuracy, proceed as follows:
  - a. Retain test setup of Figure 5-10,
- b. Repeat steps b through d of Paragraph5-22.
  - c. METER switch to AC VOLTS position.
- d. Front panel meter should indicate rms value of peak output voltage indicated on oscilloscope ( $E_0$  pk.  $\times$  0.707 =  $E_0$  rms),

### 5-24 TROUBLESHOOTING

5-25 Components within Hewlett-Packard instruments are conservatively operated to provide maximum reliability. In spite of this, parts within a unit may fail. Usually the instrument must be immediately repaired with a minimum of "down time" and a systematic approach as outlined in succeeding paragraphs can greatly simplify and speed up the repair.

### 5-26 TROUBLE ANALYSIS

- 5-27 General. Before attempting to trouble shoot this instrument, ensure that the fault is with the instrument and not with an associated circuit. The performance test (Paragraph 5-10) enables this to be determined without having to remove the instrument from the cabinet.
- 5-28 Once it is determined that the PS/A is at fault, check for obvious troubles such as an open fuse, a defective power cable, or an input power failure. Next, remove the top and bottom covers (each held by four retaining screws) and inspect for open connections, charred components, etc. If the trouble source cannot be detected by visual inspection, follow the detailed procedure outlined in succeeding paragraphs. Once the defective component has been located (by means of visual inspection or trouble analysis) correct it and reconduct the performance test. If a component is replaced, refer to the repair and replacement and adjustment and calibration paragraphs in this section.

5-29 A good understanding of the principles of operation is a helpful aid in troubleshooting, and it is recommended that the reader review Section IV of the manual before attempting to troubleshoot the unit in detail. Once the principles of operation are understood, logical application of this knowledge used in conjunction with the normal voltage readings shown on the schematic and the additional procedures given in the following paragraphs should suffice to isolate a fault to a component or small group of components. The normal voltages shown on the schematic are positioned adjacent to the applicable test points (identified by encircled numbers on the schematic and printed wiring board). Additional test procedures that will aid in isolating troubles are as follows:

- a. Reference circuit check (Paragraph 5-31). This circuit provides critical operating voltages for the instrument and faults in the circuit could affect the overall operation in many ways.
  - b. Feedback loop checks (Paragraph 5-32).
- c. Procedures for dealing with common troubles (Paragraph 5-33).
- 5-30 Note that the above mentioned procedures are applicable to both power supply and amplifier operation except in a few cases which are noted in the procedures. Generally speaking, very few

troubles that are encountered will affect the operation of the PS/A in only one of its operating modes.

- 5-31 Reference Circuit. To check the reference circuit, proceed as follows:
- a. Turn VOLTS/GAIN control fully clock-wise and set MODE switch to SUPPLY.
  - b. Turn-on unit with no load connected.
  - c. Proceed as instructed in Table 5-2.
- 5-32 Feedback Circuit. Generally, malfunction of the feedback circuit is indicated by high or low output voltages. If one of these situations occur, disconnect the load and proceed as instructed in the applicable table (5-3 through 5-6). Note that the cooling fan in Model 6824A may interfere with some voltage readings recommended in the following tables. If this occurs, the fan may be electrically disconnected in order to take the reading in question. However, the fan must be reconnected immediately after the reading has been taken, as operation of the unit without benefit of the cooling fan will cause overheating and probable damage.
- 5-33 Common Troubles. Table 5-7 provides a tabular list of the symptoms, checks, and probable causes for common troubles.

Table 5-2. Reference Circuit 'Trouble shooting

Step	Meter Common	Meter Positive	Normal Indication	Action
1	С	30	+22, 1 ±2 Vdc	If normal, proceed Step 2. If abnormal, check .N4, C12, R17, CR5, and CR6.
2	С	11	+6.2 ±0.3 Vdc	If abnormal, check VR2, R36, R56, VR5.
3	С	10	-6, 2 ±0, 3 Vdc	If abnormal, check VR1, VR3, C13, R37, R60, CR7, and CR8,

Table 5-3. High Positive Output Voltage Troubleshooting

Step	Measure	Response	Probable Cause
1	Check fuse F2	a. Blown	a. Q2, Q13, or Q14 shorted. CR21, CR22 or CR24 open.
		b. Not Blown	b. Proceed to Step 2.
2	Voltage between 23 (+) and 25 (-)	a. Less positive than +5.4V	a. CR21, or CR22, or CR24 shorted, R61 open.
	,	b. +5.4V to +5.7V	b. Check Q10 or R22 for open. Check Q1 or Q9 for short, Proceed to Step 3.
3	Voltage between C and 16	a. More negative than -60V	a. Q8 open. Q7 shorted. Q6 open.
		b56V to -60V	b. Proceed to Step 4.
4	Voltage between C and 17	a. More positive than +60V	a. Q3 open. R44 shorted.
		b. +54V to +60V	b. Check Q4 for short. Check Q11 for open, CR12 for short.

Table 5-4. High Negative Output Voltage Troubleshooting

Step	Measure	Response	Probable Cause
1	Voltage between 23 (+) and 25 (-)	a. More positive than +5.7V	a. CR21, CR22, or CR24 open.
		b. +5.4V to +5.7V	b. Proceed to Step 2.
2	Voltage between 19 and 7,8, and 9	a. OV or positive	a, Q2,Q13 or Q14 shorted, Q10 shorted.
		b. More negative than 0V	b. Proceed to Step 3.
3	Voltage between C and 16	a. Less negative than -45V	a. Q8 shorted. Q7 open. Q6 shorted.
		b45V to -50V	b. Proceed to Step 4.
4	Voltage between C and 17	a. More negative than -55V	a. Q3 shorted, R64 shorted,

Table 5-5. Low Positive Output Voltage Troubleshooting

Step	Measure	Response	Probable Cause
1	Remove Q12 from circuit	a. Normal positive output voltage	a. Q12 shorted R5 or R3 open or high resistance.
		b. Low positive output voltage	b. Reconnect Q12 and check Q1 or Q9 for open. If normal, proceed to Step 2.

Table 5-5. Low Positive Output Voltage Troubleshooting (Continued)

2 Voltage between (	C and 16 a.	Less negative than -56V	a.	Open strap between A3 -A4, Check -6,2V reference for low voltage, Check R13 for open, Check R52A for open,
	b.	-56V to -60V	b.	Proceed to Step 3,
3 Voltage between (	C and 17 a,	Less positive than +54V	a,	Check R44 or R55 for open or high resistance. Check Q4 for open.

Table 5-6. Low Negative Output Voltage Troubleshooting

Step	Measure	Response	Probable Cause
1	Remove Q5 from circuit	a. Normal negative output voltage	a. Q5 shorted. R6 or R19 open or high resistance.
		b. Low negative output voltage	b. Reconnect Q5 and check Q2,Q13, or Q14 for open. If normal proceed to Step 5
2	Voltage between C and 16	a. More negative than -50V	a. Open strap between A3-A4. Check R52A for open. Check +6.2V reference for low voltage. Check R13 for low resistance.
		b45V to -50V	b. Proceed to Step 4.
3	Voltage between C and 17	a. Less negative than -45V	a. Check R4 or R55 for open or high resistance, Check Q11 for short, Q4 for open,

Table 5-7. Common Troubles

Symptom	Checks and Probable Causes
High ripple	<ul> <li>a. Check operating setup for ground loops.</li> <li>b. Ensure that unit is not crossing over to current limiting operation under loaded conditions.</li> <li>(This will be characterized by ripple in the power supply mode and clipping of the positive and negative peaks in the amplifier mode).</li> </ul>
Oscillation (power supply mode)	a. Check feedback networks; C8-R25, C5-R24, and C7-R14.
Distortion (amplifier mode)	<ul> <li>a. Ensure that unit is not going into current limit operation.</li> <li>b. If high frequency component of output signals are distorted, check feedback networks; C8-R25, C5-R24, and C7-R14.</li> </ul>

### Table 5-7. Common Troubles (Continued)

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Poor Stability	a. Check reference voltages (Table 5-2).
<del>"</del> .	b. Noisy VOLTS/GAIN control (R52A).
	c. CR25 through CR28 leaky,
	d. Stage Q7 or Q8 defective.
Unit programs in	a. R38 or R39 open
only one direction	b. No +6. 2V or -6. 2V reverence source.
Zero Output	- 200
zero Output	a. R32 or R52A open
	b. S2 open

Excessive heat or pressure can lift the copper strip from the board. Avoid damage by using a low power soldering Iron (50 watts maximum) and following these instructions. Copper that lifts off the board should be cemented in place with a quick drying acetate base cement having good electrical insulating properties.

A break in the copper should be repaired by soldering a short length of tinned copper wire across the break,

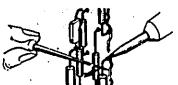
Use only high quality rosin core solder when repairing a ched circuit boards. NEVER USE PASTE FLUX. After soldering, clean off any excess flux and coat the repaired area with a high quality electrical varnish or lacquer.

When replacing components with multiple mounting pins such as tube sockets, electrolytic capacitors, and potentiometers, it will be necessary to lift each pin slightly, working around the components several times until it is free.

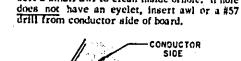
WARNING: If the specific instructions outlined in the steps below regarding etched circuit boards without eyelets are not followed, extensive damage to the etched circuit board will result.

1. Apply heat sparingly to le. d of component to be replaced. If lead of component passes through an eyelet in the circuit board, apply heat on component side of board. If lead of component does not pass through an eyelet, apply heat to conductor side of board.

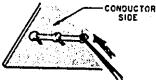
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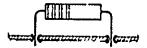
Bend clean tinned leads on new part and carefelly insert through eyelets or holes in board.

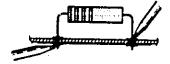


Reheat solder in vacant eyelet and quickly insert a small awl to clean inside of hole. If hole



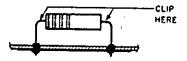
 Hold part against board (avoid overheating) and solder leads. Apply heat to component leads on correct side of board as explained in step 1.



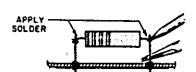


In the event that either the circuit board has been damaged or the conventional method is impractical, use method shown below. This is especially applicable for circuit boards without eyelets.

1. Clip lead as shown below,



 Bend protruding leads upward. Bend lead of new component around protruding lead. Apply solder using a pair of long nose pliers as a heat sink.



This procedure is used in the field only as an alternate means of repair. It is not used within the factory.

Figure 5-11. Servicing Printed Wiring Boards

Table 5-8: Selected Semiconductor Characteristics

Reference Designator	Characteristics	Ø Stock No.		Suggested Replacement	
Q1,4,9,10	NPN Si Power. $h_{fe} = 35 \text{ (min.)} \oplus l_{c} = 1A;$ $V_{CE} = 4V.$ $V_{CE} \text{ (sat.)} = 1V \text{ (max.)} \oplus l_{c} = 1A;$ $I_{b} = 0.1A.$ $V_{BE} = 1.5V \text{ (max.)} \oplus l_{c} = 1A;$ $V_{CE} = 4V.$ Power rating test $V_{CE} = 130V;$ $I_{c} = 500\text{mA}$ for 1 second.	1854-0250	2N3442	R. C. A.	
Q2, 13, 14	PNP Ger Power. hfe = 40 (min.) @ 8A. BVCEO = 160V. Pwr-Diss. = 150w. FT (max.) = 400 kHz.	1850-0407	B213002	Bendix	
CR5-10, 19	Si rectifier, 500ma, 200prv	1901-0026	1N3253	R.C.A.	
GR11, 12, 22, 25-28	Si rectifier 200ma, 15prv	1901-0461	1N4828	G.E.	
CR13-16	Si rectifier, 200ma, 200prv	1901-0033	1N279	Transitron	
VR3-5	Zener Diode, 22.1V ±5%, 1W.	1902-0562	1 N 3 0 2 8	Motorola	

### 5-34 REPAIR AND REPLACEMENT

5-35 Before servicing a printed wiring board, refer to Figure 5-11. Section VI of this manual contains a tabular list of the instruments replaceable parts. Before replacing a semiconductor device, refer to Table 5-8 which lists the special

characteristics of selected semiconductors. If the device to be replaced is not listed in Table 5-8, the standard manufacturers part number listed in Section VI is applicable. After replacing a semiconductor, refer to Table 5-9 for checks and adjustments that may be necessary.

Table 5-9. Checks and Adjustments After Replacement of Semiconductor Devices

Reference	Function	Check	Adjust
Q1, Q2, Q9, Q13, Q14	Regulator/Output Amplifier	Load regulation (Power Supply) Gain/frequency response (Amplifier)	
		Gurrent limit Ammeter/Voltmeter track- ing	R5 and R6 R35/R31, R43
Q3, Q4, Q6, Q11	Error and driver amplifiers	Load regulation	
Q7, Q8	Differential Amplifier	Line and load regulation Programming current	R38 and R39
Q5, Q12	Current limiting transistors	Current limit	R5 or R6
CR1-CR10	Rectifier diodes	Voltage across appro- priate filter capacitor	
VR1-VR5	Develop stable reference voltages	Positive and negative reference voltages Line regulation	

Table 5-10. Calibration Adjustment Summary

Adjustment or Calibration	Paragraph	Control Device
Meter Zero	5-38	Pointer
DC Voltmeter Tracking	541	R31
AC Voltmeter Tracking	5-42	R43
DC Ammeter Tracking	5-43	R35
Programming Current	5~45	R38 and R39
Current Limit	5-47	R5 and R6

## 5-36 ADJUSTMENT AND CALIBRATION

5-37 Adjustment and calibration may be required after performance testing, troubleshooting or repair and replacement. Perform only those adjustments that affect the operation of the faulty circuit and no others. Table 5-10 summarizes the adjustments and calibrations contained in the following paragraphs.

### 5-38 METER ZERO

- 5-39 Proceed as follows to zero meter:
- a. Turn off instrument (after it has reached normal operating temperature) and allow 30 seconds for all capacitors to discharge.
- b. Insert sharp pointed object (pen point or awl) into the small indentation near top of round black plastic disc located directly below meter face.
- c. Rotate plastic disc clockwise (cw) until meter reads zero, then rotate ccw slightly in order to free adjustment screw from meter suspension. If pointer moves, repeat steps b and c.

### 5-40 VOLTMETER TRACKING

- 5-41 To calibrate dc voltmeter tracking, proceed as follows:
- a. Connect differential voltmeter across output. Connect positive lead to common terminal on rear of PS/A.
- b. Set METER switch to VOLTS and MODE switch to SUPPLY. Adjust VOLTS/GAIN control ccw until differential voltmeter reads exactly the maximum rated negative output voltage.

- c. Adjust R31 until front panel meter also indicates maximum rated output voltage.
- 5-42 To calibrate ac voltmeter tracking, proceed as follows:
- a. Connect test setup shown in Figure 5-10 except remove load resistor  $R_{Y^*}$
- b. Set MODE switch to AMPLIFIER and oscillator frequency to 2 kHz.
- c. Adjust VOLTS/GAIN control for 100 volt peak-to-peak reading on oscilloscope.
- d. Set METER switch to AC VOLTS and adjust R43 to obtain 35.35 volt rms reading on front panel meter.

### 5-43 AMMETER TRACKING

- 5-44 To calibrate ammeter tracking proceed as follows:
  - a. Connect test setup shown on Figure 5-3.
  - b. Set METER switch to AMPS.
- c. Turn on supply and adjust VOLTS/GAIN control cw until differential voltmeter reads 1.0 Vdc.
- d. Adjust R35 until front panel meter indicates exactly the maximum rated positive output current.

### 5-45 PROGRAMMING CURRENT

5-46 To calibrate the programming current, proceed as follows:

- a. Connect differential voltmeter across output terminals.
- b. Set MODE switch to SUPPLY and turn on unit.
- c. Set VOLTS/GAIN control fully cw and FINE/ZERO control to mid-position.
- d. Adjust R38 so that differential voltmeter reads +20.2 Vdc (Model 6823A) or +52 Vdc (Model 6824A).
- e. Set polarity switch for minus input and rotate VOLTS/GAIN control fully ccw.
- f. Adjust R39 so that differential voltmeter reads -20.2 Vdc (Model 6823A) or -52 Vdc (Model 6824A).

### 5-47 CURRENT LIMIT ADJUSTMENT

5-48 To adjust the current limit so that the unit can be used to furnish maximum rated output cur-

rent, proceed as follows:

- a. Connect test setup shown in Figure 5-3 except connect variable transformer between input source and PS/A.
  - b. Short out load resistor (Ry).
- c. Set MODE switch to SUPPLY. Adjust variable transformer for 105 Vac input (low line).
- d. Turn on supply and rotate VOLTS/GAIN controls fully clockwise.
- e. Adjust R5 until differential voltmeter indicates +1.5 Vdc for Model 6823A or +1.2 Vdc for Model 6824A. Set position of input polarity switch for minus input and rotate VOLTS/GAIN control fully ccw.
- f. Adjust R6 until differential voltmeter indicates -1.5 Vdc for Model 6823A units or -1.2 Vdc for Model 6824A units.

# PARS

### SECTION VI REPLACEABLE PARTS

### INTRODUCTION

- 6-2 This section contains information for ordering replacement parts. Table 6-4 lists parts in alphanumeric order by reference designators and provides the following information:
  - a. Reference Designators. Refer to Table 6-1.
- b. Description. Refer to Table 6-2 for abbreviations.
- c. Total Quantity (TQ). Given only the first time the part number is listed except\in instruments containing many sub-modular assemblies, in which case the TQ appears the first time the part number is listed in each assembly.
  - d. Manufacturer's Part Number or Type.
- e. Manufacturer's Federal Supply Code Number. Refer to Table 6-3 for manufacturei's name and address.
  - f. Hewlett-Packard Part Number.
- Recommended Spare Parts Quantity (RS) g. for complete maintenance of one instrument during one year of isolated service.
- h. Parts not identified by a reference designator are listed at the end of Table 6-4 under Mechanical and/or Miscellaneous. The former consists of parts belonging to and grouped by individual assemblies; the latter consists of all parts not immediately associated with an assembly.

### ORDERING INFORMATION 6-3

To order a replacement part, address order or inquiry to your local Hewlett-Packard sales office (see lists at rear of this manual for addresses). Specify the following information for each part: Model, complete serial number, and any Option of special modification (J) numbers of the instrument; Hewlett-Packard part number; circuit reference designator; and description. To order a part not listed in Table 6-4, give a complete description of the part, its function, and its location.

Table 6-1. Reference Designators

A B C CB CR CR	= assembly = blower (fan) = capacitor = circuit breaker = diode = device, signal- ing (lamp)	E F J K L	= miscellaneous electronic part = fuse = jack, jumper = relay = inductor = meter
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Table 6-1. Reference Designators (Continued)

P Q R S T TB TS	= plug = transistor = resistor = switch = transformer = terminal block = thermal switch	V VR X Z	= vacuum tube, neon bulb, photocoll, etc. = zener ciode = socket = integrated circuit or network
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Table 6-2. Description Abbreviations

A = ampere	mfr = manufacturer
ac = alternating	mod, = modular or
current	modified
assy, = assembly	mtg = mounting
bd = board	$n = nano = 10^{-9}$
bkt = bracket	NC = normally closed
°C = degree	NO = normally open
Centigrade	NP = nickel-plated
cd = card	a = ohm
coef = coefficient	obd = order by
comp = composition	description
CRT = cathode-ray	OD = outside
tube	diameter
CT = center-tapped	$p = pico = 10^{-12}$
dc = direct current	P.C. = printed circuit
DPDT = double pole,	pot, = potentiometer
double throw	p-p = peak-to-peak
DPST = double pole,	ppm = parts per
single throw	million
elect = electrolytic	pvr = peak reverse
encap = encapsulated	voltage
F = farad	rect = rectifier
or = degree	rms = root mean
Farenheit	square
fxd = fixed	Si = silicon
Ge = germanium	SPDT = single pole,
H = Henry	double throw
Hz = Hertz	SPST = single pole,
IC = integrated	single throw
circuit	SS = small signal
ID = inside diameter	
ir and = incandescent	tan, = tantulum
$k = kilo = 10^3$	Ti = titanium
m = milli = 10-3	V = volt
$M = mega = 10^6$	var '= variable
$\mu = \text{micro} = 10^{-6}$	ww = wirewound
met, = metal	W = Watt

MANUAL CHANGES Model 6824A DC Power Supply Manual HP Part No. 06824-90001

Make all corrections in the manual according to errata below, then check the following table for your power supply serial number and enter any listed change (s) in the manual.

6-1

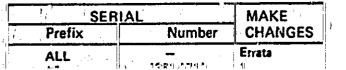


Table 6-3, Code List of Manufacturers

	<u></u>		<u> </u>
CODE NO.	MANUFACTURER ADDRESS	CODE NO,	MANUFACTURER ADDRESS
00629	EBY Sales Co., Inc. Jamaica, N.Y.	07138	Westinghouse Electric Corp.
00656		07130	
	Aerovox Corp. New Bedford, Mass.	07000	
00853	Sangamo Electric Co.	07263	Fairchild Camera and Instrument
	S. Carolina Div, Pickens, S.C.	i	Corp. Semiconductor Div.
01121	Allen Bradley Co. Milwaukee, Wis,		Mountain View, Calif.
01255	Litton Industries, Inc.	07387	Birtcher Corp., The Los Angeles, Calif.
	Beverly Hills, Calif.	07397	Sylvania Electric Prod. Inc.
01281	TRW Semicanductors, Inc.	:	Sylvania Electronic Systems
	Lawndale, Calif,	ř	Western Div. Mountain View, Calif,
01295	Texas Instruments, Inc.	07716	IRC Div. of TRW Inc. Burlington Plant
	Semiconductor-Components Div.		Burlington, Iowa
	Dallas, Texas	07910	Continental Device Corp,
01686	RCL Electronics, Inc. Manchester, N. H.		Hawthorne, Calif.
01930	Amerock Corp. Rockford, Ill.	07933	Raytheon Co. Components Div.
02107	Sparta Mfg, Co, Dover, Ohio		Semiconductor Operation
02114	Ferroxcube Corp. Saugerties, N.Y.		Mountain View, Calif.
02606	Fenwal Laboratories Morton Grove, Ill.	08484	Breeze Corporations, Inc. Union, N.J.
05660	Amphenol Corp. Broadview, Ill.	08530	Reliance Mica Corp. Brooklyn, N.Y.
02735	Radio Corp. of America, Solid State	08717	Sloan Company, The Sun Valley, Calif.
1 027 00	and Receiving Tube Div. Somerville, N.J.	08730	Vemaline Products Co. Inc. Wyckoff, N.J.
03508	G.E. Semiconductor Products Dept.	08806	General Elect, Co. Minia-
03300	· · · · · · · · · · · · · · · · · · ·	08800	
03797	Syracuse, N, Y.	onnen	ture Lamp Dept, Cleveland, Ohio
	Eldema Corp, Compton, Calif.	08863	Nylomatic Corp. Norrisville, Pa.
03877	Transitron Electronic Corp,	08919	RCH Supply Co. Vernon, Calif,
02000	Wakefield, Mass,	09021	Airco Speer Electronic Components
03888	Pyrofilm Resistor Co. Inc.	20120	Bradford, Pa,
	Cedar Knolls, N.J.	09182	*Hewlett-Packard Co, New Jersey Div,
04009	Arrow, Hart and Hegeman Electric Co.		Berkeley Heights, N. J.
	Hartford, Conn,	09213	General Elect. Co. Semiconductor
04072	ADC Electronics, Inc. Harbor City, Calif.		Prod. Dept. Buffalo, N.Y.
04213	Caddell & Burns Mfg. Co. Inc.	09214	General Elect. Co. Semiconductor
	Mineola, N.Y.		Prod. Dept. Auburn, N.Y.
04404	*Hewlett-Packard Co. Palo Alto Div,	09353	C & K Components Inc. Newton, Mass.
	Palo Alto, Calif.	09922	Burndy Corp, Norwalk, Conn.
04713	Motorola Semiconductor Prod. Inc.	11115	Wagner Electric Corp.
	Phoenix, Arizona		Tung-Sol Div, Bloomfield, N.J.
05277	Westinghouse Electric Corp.	11236	CTS of Berne, Inc. Berne, Ind.
	Semiconductor Dept, Youngwood, Pa.	11237	Chicago Telephone of Cal. Inc.
05347	Ultronix, Inc. Grand Junction, Colo.		So. Pasadena, Calif.
05820	Wakefield Engr. Inc. Wakefield, Mass.	11502	IRC Div. of TRW Inc. Boone Plant
06001	General Elect, Co. Electronic	ŀ	Boone, N.C.
I	Capacitor & Battery Dept, Irmo, S.C.	11711	General Instrument Corp
06004	Bassik Div. Stewart-Warner Corp.		Rectifier Div. Newark, N.J.
	Bridgeport, Conn.	12,36	Philadelphia Handle Co. Inc.
06486	IRC Div. of TRW Inc.	•	Camden, N.J.
	Semiconductor Plant Lynn, Mass.	12615	U.S. Terminals, Inc. Cincinnati, Ohio
06540	Amatom Electronic Hardware Co. Inc.	12617	Hamlin Inc. Lake Mills, Wisconsin
	New Rochelle, N.Y.	12697	Clarostat Mfg, Co, Inc. Dover, N. H.
06555	Beede Electrical Instrument Co.	13103	Thermalloy Co. Dallas, Texas
	Penacook, N.H.	14493	*Hewlett-Packard Co. Loveland Div.
06666	General Devices Co. Inc.		Loveland, Colo.
	Indianapolis, Ind.	14655	Cornell-Dubilier Electronics Div.
06751	Semcor Div. Components, Inc.	1 - 13-3	Federal Pacific Electric Co.
	Phoenix, Arizona	ĺ	Newark, N. J.
06776	Robinson Nugent, Inc. New Albany, Ind.	14936	General Instrument Corp. Semicon-
06812	Torrington Mig. Co., West Div.	1 1 3 3 0	ductor Prod. Group Hicksville, N. Y.
******	Van Nuys, Calif.	15801	
07137	Transistor Electronics Corp.	16299	Fenwal Elect. Framingham, Mass. Corning Glass Works, Electronic
" " " "	Minneapolis, Minn,	10233	_
	windeapous, winn,	<u></u>	Components Div. Raleigh, N.C.
		-	

<sup>\*</sup>Use Code 28480 assigned to Hewlett-Packard Co., Palo Alto, California

Table 6-3. Code List of Manufacturers (Continued)

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NO.	MANUFACTURER ADDRESS	COL	1 / 1 \
16758	manufacture manufacture of the contract in order and the contract in o	7050	0111011 01111
17545		7090 7090	Belden Corp. Chicago, Ill.
17803	I remember Comercia and managing the COLD	7121	
	Semiconductor Div, Transducer Plant Mountain View, Calif.	7140	Cambridge, Mass.
17870	Daven Div. Thomas A. Edison Industries McGraw-Edison Co. Orange, N. J.	7145	Edison Co. St. Louis, Mo.
18324	Signetics Corp. Sunnyvale, Calif.	7146	
19315	Bendix Corp. The Navigation and Control Div. Teterboro, N. J.	7159	Los Angeles, Calif.
19701	Electra/Midland Corp. Mineral Wells, Texas	7170	Centralab Div. Milwaukee, Wis.
21520	Fansteel Metallurgical Corp.	7170	Wire Co. Div. Williamstown, Mass.
22229	No. Chicago, Ill, Union Carbide Corp. Electronics Div.	7170 7174	
22753	Mountain View, Calif,		Chicago, Ill.
23936	UID Electronics Corp. Hollywood, Fla. Pamotor, Inc. Panpa, Texas	7178	5 Ginch Mfg, Go, and Howard  B. Jones Div. Chicago, III.
24446 24455	Pamotor, Inc. Pampa, Texas General Electric Co. Schenectady, N.Y. General Electric Co. Lamp Div. of Con-	7198	4 Dow Corning Corp. Midland, Mich.
	sumer Prod, Group	7213	Willimantic, Conn.
24655	Nela Park, Cleveland, Ohio General Radio Co. West Concord, Mass.	72619 7269	Dialight Corp. Brooklyn, N.Y.
24681	LTV Electrosystems Inc Memcor/Com-	7276	Drake Mfg. Co. Harwood Heights, Ill.
26982	ponents Operations Huntington, Ind. Dynacool Mfg. Co. Inc. Saugerties, N.Y.	7296	Elastic Stop Nut Div, of
27014	National Semiconductor Corp.	72982	Erie Technological Products Inc. Erie, Pa.
28480	Santa Clara, Calif. Hewlett-Packard Co. Palo Alto, Calif.	73096	indiction Colling
28520 28875	Heyman Mfg. Co. Kenilworth, N.J. IMC Magnetics Corp.	721.66	Helipot Div. Fullerton, Calif.
1 .	New Hampshire Div. Rochester, N. H.	73168	
31514	SAE Advance Packaging, Inc. Santa Ana, Calif.	73445	Dynamics Div. Torrance, Calif.
31827 33173	Budwig Mfg. Co. Ramona, Calif.	1	Hicksville, N. Y.
35434	G. E. Co. Tube Dept. Owensboro, Ky. Lectrohm, Inc. Chicago, Ill.	73506	Bradley Semiconductor Corp.  New Haven, Conn.
37942	P.R. Mallory & Co. Inc.	73559	Carling Electric, Inc. Hartford, Conn.
42190	Indianapolis, Ind, Muter Co, Chicago, Ill.	73734	Federal Screw Products, Inc. Chicago, Ill.
43334	New Departure-Hyatt Bearings Div. General Motors Corp. Sandusky, Ohio	74193 74545	Heinemann Electric Co. Trenton, N. J.
44655	Ohmite Manufacturing Co. Skokie, Ill.	74868	
46384	Penn Engr, and Mfg. Corp, Doylestown, Pa.	74970	Danbury, Conn,
47904 49956	Polaroid Corp. Cambridge, Mass.	75042	IRC Div. of TRW, Inc. Philadelphia, Pa.
55026	Raytheon Co Lexington, Mass. Simpson Electric Co. Div. of American	75183	*Howard B, Jones Div. of Cinch
56289	Gage and Machine Co. Chicago, Ill.	75376	Kurz and Kasch, Inc. Dayton, Ohio
58474	Sprague Electric Co, North Adams, Mass. Superior Electric Co. Bristol, Conn.	75382 75915	Kilka Electric Corp. Mt. Vernon, N.Y. Littlefuse, Inc. Des Plaines, Ill.
58849	Syntron Div, of FMC Corp.  Homer City, Pa.	76381	Minnesota Mining and Mfg. Co.
59730	Thomas and Betts Co. Philadelphia, Pa.	76385	St. Paul, Minn. Minor Rubber Co. Inc. Bloomfield, N.J.
61637 63743	Union Carbide Corp. New York, N.Y. Ward Leonard Electric Co.	76487	James Millen Mfg. Co. Inc.
	Mt, Vernon, N.Y.	76493	J. W. Milier Co. Malden, Mass. Compton, Calif.
*IIco Co	ndo 71705	<u> </u>	

CODE NO.	MANUFACTURER ADDRESS
76530	Cinch City of Industry, Calif.
76854	Gak Mfg. Co. Div. of Oak
	Electro/Netics Corp. Crystal Lake, Ill
77068	Bendix Corp., Electrodynamics Div.
77100	No. Hollywood, Calif
77122 77147	Palnut Co. Mountainside, N. J.
77221	Patton-MacGuyer Co. Providence, R. I. Phaostron Instrument and Electronic Co.
,,,,,,	South Pasadena, Calif.
77252	Philadelphia Stael and Wire Corp.
	Philadelphia, Pa.
77342	American Machine and Foundry Co.
	Potter and Brumfield Div. Princeton, Ind.
77630	TRW Electronic Components Div.
77764	Camden, N.J. Resistance Products Co. Harrisburg, Pa.
78189	Illinois Tool Works Inc. Shakeproof Div.
/0103	Elgin, Ill.
78452	Everlock Chicago, Inc. Chicago, III.
78488	Stackpole Carbon Co. St. Marys, Pa.
78526	Everlock Chicago, Inc. Chicago, Ill. Stackpole Carbon Co. St. Marys, Pa. Stanwyck Winding Div. San Fernando
	Electric Mig. Co. Inc. Newburgh, N.Y.
78553	Tinnerman Product:, Inc. Cleveland, Ohio
78584	Stewart Stamping Corp, Yonkers, N.Y.
79136 79307	Stewart Stamping Corp, Yonkers, N.Y. Waldes Kohincor, Inc. Whitehead Metals Inc.  Yonkers, N.Y. L.I.C., N.Y. New York, N.Y.
79727	Continental-Wirt Electronics Corp.
, 3, 2,	Philadelphia, Pa
79963	Philadelphia, Pa. Zierick Mfg. Co. Mt. Kisco, N.Y.
80031	Mepco Div. of Sessions Clock Co.
	Morristown, N.J.
80294	Bourns, Inc. Riverside, Calif.
81042	Howard Industries Div. of Msl Ind. Inc.
81073	Racine, Wisc, Grayhill, Inc. La Grange, Ill.
81483	Grayhil, Inc. La Grange, Ill, International Rechifier Corp.
	El Segundo, Calif.
81751	Columbus Electronics Corp. Yonkers, N.Y.
'82099	Goodyear Sundries & Mechanical Co. Inc.
	New York, N, Y,
82142	Airco Speer Electronic Components
82219	Du Bois, Pa. Sylvania Electric Products Inc.
VLL.3	Electronic Tube Div. Receiving
	Tube Operations Emporium, Pa.
82389	Switchcraft, Inc. Chicago, Ill.
82647	Metals and Controls Inc. Control
	Products Group Attleboro, Mass.
82866	Research Products Corp. Madison, Wis.
82877 82893	Rotron Inc. Woodstock, N.Y. Vector Electronic Co. Glendale, Calif.
83058	Vector Electronic Co.  Carr Fastener Co.  Glendale, Calif.  Cambridge, Mass.
83186	Victory Engineering Corp.
	Springfield, N. J.
83298	Bendix Corp, Electric Power Div,
	Eatontown, N.J.
83330	Herman H. Smith, Inc. Brooklyn, N.Y.
83385	Central Screw Co. Chicago, Ill.
83501	Gavitt Wire and Cable Div. of
<u></u>	Amerace Esna Corp. Brookfield, Mass.

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CODE	MANUFACTURER ADDRESS
83508	Grant Pulley and Hardware Co, West Nyack, N.Y.
83594	Burroughs Corp. Electronic
83835 83877	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
84171 84411 86684	Arco Electronics, Inc. Great Neck, N.Y. TRW Capacitor Div. Ogaliala, Neb, RCA Corp. Electronic Components
86838 87034	
87216 87585	Philco Corp, Lansdale Div, Lansdale, Pa. Stockwell Rubber Co. Inc.
87929 88140	Philadelphia, Pa, Tower-Olschan Corp. Bridgeport, Conn. Cutler-Hammer Inc. Power Distribution and Control Div. Lincoln Plant
88245	Lincoln, Ill. Litton Precision Products Inc, USECO Div. Litton Industries Van Nuys, Calif.
90634	Gulton Industries Inc. Metuchen, N. J. United-Car Inc. Chicago, Ill.
91345	Miller Dial and Nameplate Co.  El Monte, Calif.
91506 91637	
91662 91929	Elco Corp. Willow Grove, Pa. Honeywell Inc. Div. Micro Switch
92825 93332	Freeport, Ill. Whitso, Inc. Schiller Pk., Ill. Sylvania Electric Prod. Inc. Semi-
93410	conductor Prod. Div. Woburn, Mass. Essex Wire Corp. Stemco
94144	Controls Div, Mansfield, Ohio
94154	Ind. Components Oper. Quincy, Mass. Wagner Electric Corp. Tung-Sol Div. Livingston, N. I.
94222 95263	Tung-Sol Div.  Southco Inc.  Lester, Pa.  Lecraft Mfg. Co. Inc.  L. I. C., N. Y.
95354 95712	Methode Mfg, Co. Rolling Meadows, Ill. Bendix Corp. Microwave
95987 96791	Devices Div. Franklin, Ind. Weckesser Co. Inc. Chicago, Ill.
97464	Amphenol Corp. Amphenol Controls Div. Industrial Retaining Ring Co.
97702	Irvington, N. J. IMC Magnetics Corp. Eastern Div.
98291 98410	Westbury, N.Y. Sealectro Corp. Mamaroneck, N.Y. E'fC Inc. Cleveland. Ohio
98978	E'IC Inc. Cleveland, Ohio International Electronic Research Corp.  Burbank, Calif.
99934	Renbrandt, Inc. Boston, Mass.

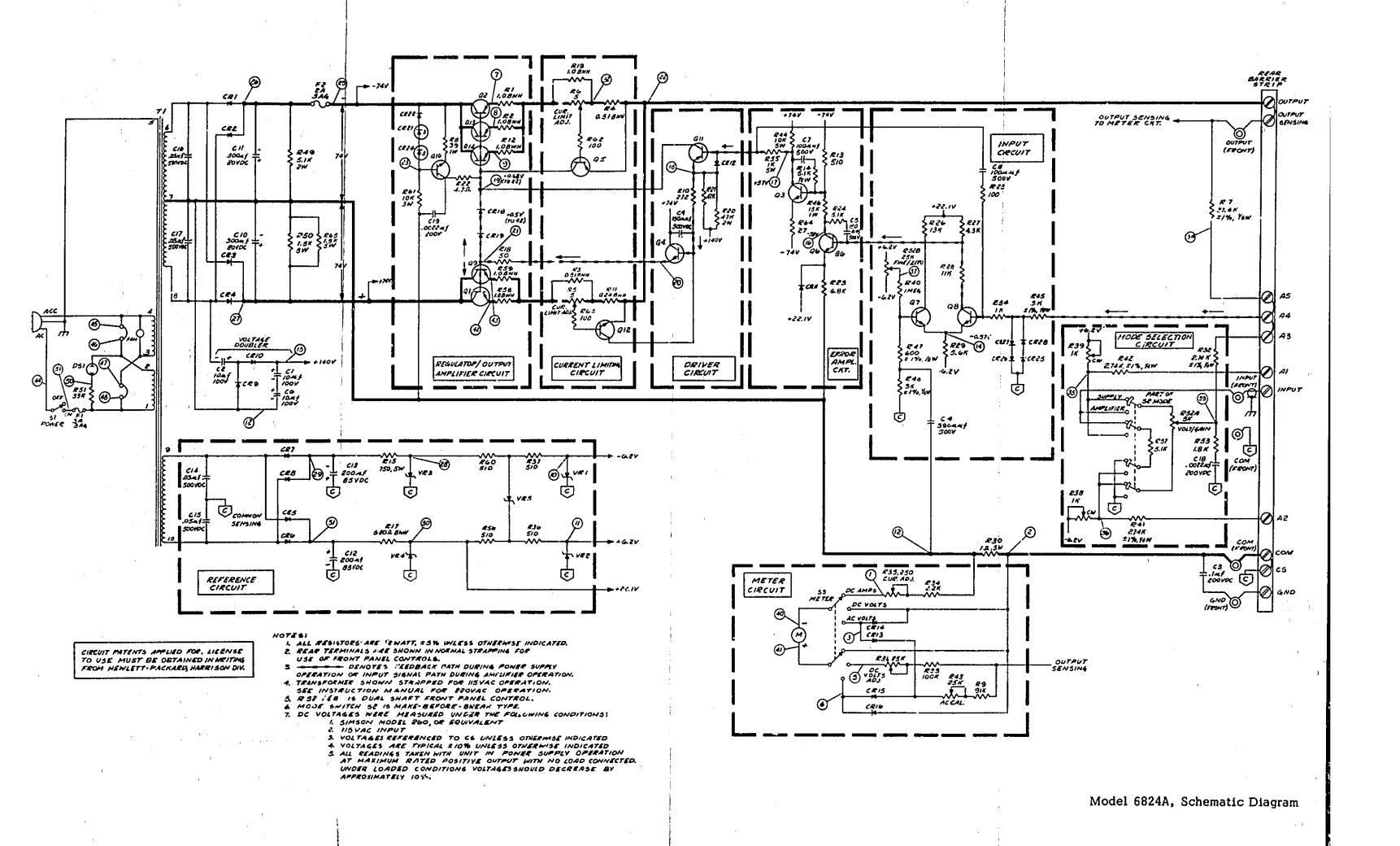
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1							
Reference	•		Mfr. Part #		Mfr.	<b>€</b>	
Designator	Description	Quantity		Mfr.	Code		
		<u> </u>	U. Type	IVIII.	Code	Stock No.	<u>RS</u>
C1, 2, 6	fxd, elect 10µf 100vdc	3	30D106F100 DC	2 Cmma			
C3	fxd, film , luf 200vdc	ì	192P10492		56289	0180-0091	1
C4	fxd, Mica 390µµf 500vdc	i		Sprague	56289	0160-0168	1
C5,9	fxd, Mica 150µµf 500vdc		RCM15E391J	Arco	04062	0140-0037	1
C1,8	fxd, Mica 100 upf 500 vdc	2	RCM15E151J	Arco	04062	0140-0067	1
C10, 11	fud along 2000 500 to	2	RCM15E101J	Arco	04062	0140-0041	1
C12, 13	fxd, elect 3000µf 80vdc	2	D42492	HLAB	09182	0180-1987	1
	fxd, elect 200µf 85vdc	2	D31642	HLAB	09182	0180-1854	ĩ
C14, 15, 16,							-
17	fxd, ceramic . 05µf 500vdc	4	33C17A	Sprague	56299	0150-0052	1
C18,19	fxd, film .0022µf 200vdc	- 2	192P22292	Sprague	56289	0160-0154	1
				~,,~,=	0000	0100-0104	1
CR1-4	Rect. si. 3A 200prv	4	MR1032B	Motorola	04713	1901-0416	
CR5-10, 19	Rect. si. 500ma 200prv	7	1N3253	R, C, A.	02735		4
CR11, 12, 22	•	·	***************************************	K, O, A,	02/33	1901-0389	6
25-28	Rect. si. 200ma 15prv	7		*** • •			
CR13-16	Rect. si. 200ma 200prv	4		HLAB	09182	1901-0461	6
CR18	Rect. si. 70V 100prv		13703	HLAB	09182	1901-0033	4
CR21, 24	Rect. si. 2.4V 150ma	1	1N91	G. E.	03508	1911-0001	1
OREX, ET	Rect. St. 2.49 150ma	2		HLAB	09182	1901-0460	2
DS1	T						
D01	Lamp, Neon part of S1 ass	y Ref.		HLAB	09182	2140-0244	1
***							-
F1	Fuse Cartridge (AC)						
	3A @ 250V 3AG	1	312003	Littlefuse	75913	2110-0003	5
F2	Fuse Cartridge (DC)				, , , , ,	2110-0003	3
	2A @ 250V 3AG	1	312002	Littlefuse	75012	2110 0000	-
				PittleIdse	73313	2110-0002	5
Q1, 4, 9, 10	Power NPN si.	4		HLAB	00100	1054 0050	
Q2, 13, 14	Power PNP Ger.	3			09182	1854-0250	4
Q3	SS NPN si.	1	40.400	HLAB	09182	1850-0407	3
Q5,11	SS PNP si.		40422	R. C. A.	02735	1854-0253	1
Q6	SS PNP si. SELECTED for	2	2N2907A	Sprague	56289	1853-0099	2
40		_					
07	BVCEL > 100V	1	40362	R. C. A.	02735	1853-0041	1
Q7,:,,12	SS NPN si.	3	2N3417	G. E.	03508	1854-0087	3
						2001 0007	3
R1, 2, 12,							
19,58,59	fxd, ww 1. $0x \pm 5\%$	6	Type BWH	I.R.C.	07716	0811-1672	•
R3, 4	fxd, ww .51 <sub>A</sub> ±5%	2	Type BWH	I, R. C.			2
R5,6	var. ww 5 <sub>A</sub>	2	Type 110-F4		07716	0811-0929	1
R7	fxd, film 27.4K ±1% 1/8w	ī	Type MF5C T-O	C. T. S.	11236	2100-1821	1
RB	fxd, ww 39, ±5%	ī	Type BWH	Electra	19709	0757-0452	1
R9	fxd, comp 91K, ±5% \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	î	_*-	I; R, C,	07716	0811-2244	1
R10,64	fxd, comp 27, ±5% ½w	2	EB-9135	A, B,	01121	0686-9135	1
R11	fxd. ww . 24 <sub>h</sub> ±5%		EB-2705	A, B,	01121	0686-2705	1
R13, 36, 37,	120. 111 . 236 236	1	Type BWH	I.R.C.	07716	0811-1758	1
56,60	fud some \$10 years	_					
R14, 24, 57	fxd, comp 510 <sub>A</sub> ±5% ½w	5	EB-5115	A, B.	01121	0686-5115	1
	fxd, comp $5.1K_1 \pm 5\% \frac{1}{2}W$	3	EB-5125	A. B.	01121	0686-5125	ī
R15, 17, 55	fxd. ww 1K ±5% 5w	. 3	Type 5XM		63743	0812-0099	î
R16	NOT ASSIGNED	-	-	_		-	•
R18	fxd, comp 50a ±5% 5w	1	243E5005	Sprague	56289	0811-1854	
R20, 21	fxd, met. ox 47Kn ±5% 2w	2	Type C42S		16299	0764-0031	1
R22	fxd, comp 4.7 <sub>A</sub> ±5% ½w		Type EB	_	01121		1
R23	fxd, comp $6.8K_{\Lambda} \pm 5\% \frac{1}{2}W$		EB-6825	_		0698-0001	1
R25,62,63	fxd, comp 1000 ±5% 3w		EB-1015	_	01121	0686-6825	1
R26	fxd, comp 13K, ±5% 1w		EB-1335	_	01121	0686-1015	1
R27	fxd, comp 4.3K, ±5% \u00e4w				01121	0686-1335	1
R28	fxd, comp 11K <sub>1</sub> ±5% ±w		EB-4325		01121	0686-4325	1
	fxd, comp 5,6K, ±5% ½w	_	EB-1135	_	01121	0686-1135	l
-		1	EB-5625	A. B.	01121	0686-5625	1

Reference			Mfr. Part #		Mfr.	<b>₽</b>	
Designator	Description Q	uantity		Mfr.	Code	Stock No.	RS
20013							
R30	fxd. ww la ±5% 5w	1	Type 5XM	W.L.	63743	0811-1340	1
R31,43	var. comp.25Ka Series 70	2			09182	2100-1534	1
R32,41,42	fxd, met. film 2.74Ka $\pm 1\% \frac{1}{4}$	w 3	Type MF6-CT-O		19707	0757-0742	1
R33	fxd, comp 100Ka ±5% ½w	1	EB-1045		01121	0686-1045	1
R34	fxd, comp 2.2Ka $\pm 5\% \frac{1}{2}$ w	1	EB-2225		01121	0686-2225	1
R35	var. ww 250 <sub>5</sub>	1	Type 110-F4		11236	2100-0439	1
R38,39	var. ww lKn	2	Type 110-F4		11236	2100-0391	1
R40	fxd, conp 1 MEGa ±5% \(\frac{1}{2}\text{w}\)	1	EB-1055	•	01121	0686-1055	1
R44	fxd. ww 10Ka ±1% 5w	1			09182	0811-1866	1
R45	fxd, met. film $3K_h \pm 1\% 1/8w$	1	Type CEA T-O		07716	0757-1393	1
R46	fxd, comp 15Kn ±5% lw	1	GB-1535	_	01121	0689-1535	1
R47	fxd, met. film $600a \pm 1\% 1/8v$	v l	Type CEA T-O		07716	0757-1100	1
R48	fxd, met. film $3.01$ Ka $\pm 1\%$ $\frac{1}{4}$ V	v l	Type MF6-C T-O		19709	0757-0339	1
R49	fxd, met. ox $5.1$ K $_{\Delta} \pm 5\%$ 2w	1	Type C42S		16299	0698-3644	1
R50,65	fxd. ww 1.5K ±5% 5w	2	243E1525		56289	0811-1863	1
R51	fxd, comp 33Kn $\pm$ 5% $\frac{1}{2}$ w	1	EB-3335		01121	0686-3335	1
R52	var. ww DUAL 5K -25K	1		HLAB	09182	2100-1805	1
R53	fxd, comp 1.8K $_{\Lambda} \pm 5\% \frac{1}{2}$ w	1	EB-1825		01121	0686-1825	1
R54	fxd, comp lKn ±5% ½w	1	EB-1025		01121	0686-1025	1
R61	fxd. ww 10K ±5% 3w	1	242E1035	Sprague	56289	0811-1816	1
<b>~</b> 3	Guidant (Dillon Files) ON (OPE	,	54-61681 - 26 AlH	Oak	87034	3101-0100	1
S1	Switch (Pilot Lite) ON/OFF		34-01001 - 20 MIN		09182	3100-1917	î
S2	Switch, Wafer Mode	1 1			09182	3100-1916	î
S3	Switch, Wafer-Meter	•		IIII	03102	0100 1010	-
Tl	Power Transformer	1		HLAB	09182	9100-1895	1
VR1,2	Diode zener 6 2V ±5%	2	1N821	N.A.	06486	1902-0761	2
VR3,4,5	Diode zener 22.1V ±5% 1w	3			09182	1902-0562	3
	Chassis-side formed (right)	1			09182	06824-00001	
	Chassis-side formed (left)	1	•		09182	06824-00002	
	Panel - front	1			09182	06824-00003	
	Chassis - rear	1		HLAB	09182	06824-00004	
	Heat sink	1		HLAB	09182	06824-20002	
	Bracket - P.C. Board	2		HLAB	09182	06824-00005	
	Cover	2		HLAB	09182	5000-6092	
	Angle Guard	1		HLAB	09182	06824-20001	_
•	5 Way binding post (maroon	) 2	DF21 Mn.	HLAB	09182	1510-0040	1
	5 Way binding post (black)	3	DF 21 BC	Superior	58474	1510-0039	1
	Line cord plug PH151 $7\frac{1}{2}$ ft.	1	KH-4096	Beldon	70903	8120-0050	1
	Strain relief bushing	1	SR-5P-l	Heyco	28520	0400-0013	1
	Knob 🛊 insert pointer	2		HLAB	09182	0370-0084	1
	Knob 17/64 insert pointer	1		HLAB	09182	0370-0101	1
	Knob 3/16	1		HLAB	09182	0370-0179	1
	Jumper	4	422-13-11013	Cinch	71785	0360-1143	1
	Barrier strip	1		HLAB	09182	0360-1285	1
	Rubber bumper	4	MB50	Stockwell		0403-0088	1
	Rubber bumper	6	4072	Stockwell	87575	0403-0086	1
	Delrin standoff ½ Dia.						
	5/8 long 8-32 tap	3		HLAB	09182	0380-0701	1
	Phenolic standoff 1 Dia.						
	5/8 long 6-32 thread	2	8694-PH 0632	Amatom	06540	0380-0702	1
	Bezel 1 MOD	1		HLAB	09182	4040-0294	1
	Fuse holder	2	342014	Littlefuse		1400-0084	1
	Lockwasher	1	1234-08	Shakeproof		2190-0037	1
	Nut	1	903-12	Littlefuse	75915	2950-0038	1

Reference Designator	Description	Quantity	Mfr. Part# or Type	Mfr.	Mfr. Code	® Stock No.	RS
1	Mates 17 /p 21 u = t = c						<u> </u>
1	Meter H/P 3½" size						. !
	0-60V 0-1.2A zero center	r 1		HLAB	09182	1120-1124	'n
	Spring	4		HLAB	09182	1460-0256	2
	Fan Motor	1		HLAB	09182	3140-0010	ĩ
	Fan Blade 4 inch dia.	1	4 RHB 183B	P. M. Motor		3160-0028	1
	Fastener	10	C8091632-24B				1
	Captive nut	10	CLA-832-2	Tinnerman		0510-0275	2
	Spacer (br. cad. pl)			Penn. Eng.		0590-0395	5
		2	#6 x 1 Lg	HLAB	09182	0380-0716	.1
	Swage nut (N1 Pl						
	6-32 x 3/16 Lg)	4	110211	Fed. Scr	73734	0380-0390	1
	Mica insulator	2	734	Reliance	08530	0340-0174	- 1
	Insulator, Transistor pin	10			09182	0340-0166	2
	Insulator	10			09182		
	Transistor Socket	5				0340-0168	2
		J		HLAB	09182	1200-0044	5



# MANUAL CHANGES

### MANUAL CHANGES Model 6824A DC Power Supply Manual HP Part No. 06824-90001

Make all corrections in the manual according to errata below, then check the following table for your power supply serial number and enter any listed change (s) in the manual.

		MAKE				
	SERIAL					
Prefix	Number	CHANGES				
ALL	-	Errata				
6F.	0361-0710	1				
6F	0711-0760	1,2				
6F	0761-0880	1,2,3				
6F	0881-1060	1,2,3,4,				
6F	1061-1100	1 thru 6				
6F	1101-1140	1 thru 6				
1026A	1105, 1123	1 thru 6, and 9				
1	1128, 1135,	:				
1 12	1136, 1137					
OC	1141-1180	1 thru 7				
1026A	1181-1220	1 thru 8				
1027A	1221-1360	1 thru 9 1 thru 10				
1027A	1361-1540	1 thru 11				
1027A	1541-1570 1571-1660	1 thru 12				
1027A	1661-2340	1 thru 13				
1027A	2341-2400	1 thru 14				
1027A 1706A	2401-2430	1 thru 15				
1716A	2431-2700	1 thru 16				
1914A	2701-2709	1 thru 18				
1914A 19	2710-2730	1 thru 17				
1932A	2731-2790	1 thru 18				
1946A	2791-2820	1 thru 19				
2001A	2821- 3190	1 thru 20,21*				
2205A	3191-3220	1-21,22†				
2227A	3221-up	1-22				
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### ERRATA:

In the replaceable parts list and on the schematic diagram, make the following changes:

C5, 11: Change to SS PNP Si., HP Part No. 1853-0099.

R10: Change value on schematic to  $27\Omega$ . Value in parts list is correct.

R1, 2, 12, 19, 58, 59: Change to HP Part No. 0811-1666. Value remains unchanged.

Rubber Bumper: Change total quantity from 6 to 4. In Paragraph 1-13, add Option 07: Finer Volts Gain

Control. Allows greater accuracy in setting the supply/amplifier output. The option consists of changes in the following parts (1 each):

R52A (Volts Gain): 5kΩ, ±5%, 19-turn HP Part No.

2100-1865.

R52B (Fine Zero): 25kΩ,±5%, HP Part No. 2100-1862.

Knob: HP Part No. 0370-0137.

Modified Front Panel: F'P Part No. 06824-60002. Under OUTPUT IMPEDANCE on Page 1-3, change the last sentence to read as follows:

"Less than 3 $\Omega$  from 1kHz to 10kHz and less than 30 $\Omega$ from 10kHz to 100kHz."

On Page 5-3, change the caption of Figure 5-2 to read: Differential Voltmeter Substitute Test Setup.

In Table 1-1, make the following changes:

PROGRAMMING SPEED:

Less than 50µsec are required to program between -25V 1+25V. Less than 17µsec are required to program between 10% and 90% of the half-voltage span.

Less than 125usec are required to program between -50V and +50V. Less than 25µsec are required to program between 10% and 90% of the maximum, voltage span. FREQUENCY RESPONSE:

DC to 20kHz (±3dB) at half-rated output (±25V, 50V

DC to 10kHz (±3dB) at full output.

DISTORTION:

Output distortion is less than 0.02% at 1kHz and halfrated output.

MAXIMUM PHASE SHIFT (at Half-Rated Output): ' (No further changes.)

### **CHANGE 1:**

In the replaceable parts table, make the following changes: Q6: Change to SS PNP Si., SS2109, Motorola, HP Part No. 1853-0037.

Fan Motor: Change to Howard 1175-06-4155, HP Part No. 3140-0406.

Fan Blade: Change to Howard 6-123-125, HP Part No. 3160-0098.

### CHANGE 2:

In the replaceable parts table make the following changes: C10,11: Change to HP Part No. 0180-2193. Clamp (C10,C11): Add 2, HP Part No. 0160-2149.

### CHANGE 3:

In Section II of the manual, add the following note to the instructions for rack mounting two units:

### NOTE

When rack mounting two units, attach the sides of the units together using the 6-32 screw (supplied with the 14525A rack kit) and the threaded insert nut attached to the right side of each unit under the heat sink. Disregard the 6-32 hex nut/lockwasher supplied with the rack mounting kit.

### CHANGE 4:

In the replaceable parts table make the following changes: Q3: Change to 2N4240, RCA, HP Part No. 1854-0311. CHANGE 5:

In the replaceable parts table and on the schematic

diagram, make the following changes: R20, R21: Change to 36k, 2W, ±5% HP Part No. 0698-3651.

### CHANGE 6:

In the replaceable parts table and on the schematic diagram, make the following changes:
R35: Change to 500 ohms, HP Part No. 2100-0328.

### CHANGE 7:

The serial number prefix of the instrument has been changed from "6F" to "0C,"

In the replaceable parts table and on the schematic dicgram, make the following changes:

R40: Change to 510k, HP Part No. 0686-5145, CR11: Change to HP Part No. 1901-0460.

### CHANGE 8:

In the replaceable parts list and on the schematic diagram, make the following changes:

R38, 39: Change to 390A, ±5%, 3W, HP Part No. 0811-1799.

R66, 67: Add, selected value,  $\pm 1\%$ , 1/8W, connected in parallel with R38 and R39 respectively. R32: Change to 2.67ka,  $\pm 1\%$ ,  $\frac{1}{4}W$ , HP Part No. 0811-1941.

R27: Change to 15kn, ±1%, 1/8W, HP Part No. 0757-0446.

R28: Delete and replace with jumper.

R24: Change to nominal value  $100_A$ ,  $\pm 5\%$ ,  $\frac{1}{2}$ W, HP Part No. 0 95-1015.

C5: Change to 510pF, 500Vdc, HP Part No. 0140+0047.

CR11: Replace with zener diode (anode connected to emitter of Q6) VR6, 6,19V, 400mW, HP Part No. 1902-0049.

CR29: Add, cathode connected to emitter of Q6 and anode connected to base of Q6; 15prv, 400mW, HP Part No. 1901-0461.

Q7,8: Change to single-housing dual differential amplifier, HP Part No. 1854-0221.

VR1, 2: Change to 6.2V, ±5%, 250mW, HP Part No. 1902-1221,

R23: Change to 5.6ka, ±5%, ½W, HP Part No. 0686-5625.

F2: Replace with short circuit.

Fuse Holder, HP Part No. 1400-0084: Delete.

Lockwasher, HP Part No. 2190-0037: Delete, Nut, HP Part No. 2950-0038: Delete. Chassis, rear: Change to HP Part No. 06824-00007.

Serial ID Plate: Add, HP Part No. 7120-1111.

### CHANGE 9:

In the replaceable parts list, make the following change:

P. C. Board Assembly: Add HP Part No. 06824-60021. Assembly omitted from original parts list in manual; this new P. C. Board Assembly number reflects changes 1 through 9 inclusive. Fan Motor: Change to HP Part No. 3140-0745.

### CHANGE 10:

In the replaceable parts list and on the schematic diagram make the following changes:

R20, 21: Change to 15km, ±5%, 5W, HP Part No. 0811-1867.

S1: Change to new type pushbutton switch, HP Part No. 3101-1248.

R61: Change to  $10k_{\Lambda}$ ,  $\pm 5\%$ , 1W, HP Part No. 0689-1035.

### CHANGE 11:

In the replaceable parts list and on the schematic, make the following change:

C6: Change to 56pF, HP Part No. 0140-0191.

### CHANGE 12.

The standard colors for this instrument are now mint gray (for front and rear panels) and olive gray (for all top, bottom, side, and other external surfaces). Option X95 designates use of the former color scheme of light gray and blue gray. Option A85 designates use of a light gray front panel with olive gray used for all other external surfaces. New part numbers are shown below.

### CHANGE 13.

In the replaceable parts list and on the schematic, make the following change:

R33: Change to 110KA ± 1%, 1/8W, HP Part No. 0757-6466.

DESCRIPTION	HP PART NO.				
DESCRIPTION	STANDARD	OPTION A95	OPTION X95		
Front Panel	06824-00010	06824-00003	-		
Chassis, Right Side	06824-00008	◆	06824-0000		
Chassis, Left Side	-06824-00009	<b>4</b>	'   06824-0000;		
Chassis, Rear	06824-00011	<b>——</b>	06824-0000		
Cover	5000-9419	<b></b>	5000-6016		
Rack Kit, 1 unit (accy)	14515A	14515A-A85	4		
Rack Kit, 2 units "	14525A	145 -A85			

### ERRATA:

Add to the parts list the replacement lamp for illuminated switch 3101-1248, which is used in those supplies that include Change 10. The HP Part No. of the type AlH lamp is 2140-0244.

### CHANGE 14:

Add to paragraph 2-18 and to the parts list the correct fuse (F1) for Option 028 (230Vac) instruments. Fuse F1 should be a 1A slow-blow type HP Part No. 2110-0007.

### ERRATA:

Effective January 1st, 1977, Option 007 (10-turn volts-gain control) has been redesignated Option 009. Make this change wherever Option 007 is mentioned in the manual.

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### . CHANGE 15:

Add to the parts list on page 6-7 one fuseholder spacer, HP Part No. 5020-2568. This plastic spacer had been added to fuseholder 1400-0084 outlisde the rear panel to increase the distance between primary and secondary circuits for increased safety.

### CHANGE 16:

On page 6-6, and on the schematic, change the resistanvalue for R31 and R43 (in the meter circuit) to 20k±1 change their HP Stock No. to 2100-0558.

### **CHANGE 17:**

On page 6-6, change the HP part numbers for the front panel + and — binding posts to: 1510-0091, red, qty 2. Change the HP part number for the front panel GND binding post to 1510-0107, black, qty 1. Add the HP part number for nylon nuts: 2950-0144, qty 3. 4-9-79

### CHANGE 18:

In the replaceable parts list and on the schematic, delete Q2 and R1. Change Q13 and Q14 to silicon trnasistors, HP Part No. 1853-0485; change CR18 to silicon, 1901-0327; change R8 to 30 ohm  $\pm 5\%$ , 2W, 0764-0041; change R61 to 7.5k  $\pm 5\%$ , 3W, 0811-1815.

### **CHANGE 19:**

In the replaceable parts list, change switch S1 to HP Part No. 3101-2287. On the schematic diagram, change switch S1 to show two switch sections, one which breaks the AC line (as shown).

The other breaks the ACC line.

### CHANGE 20:

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In the replaceable parts list, change C3 to 0.1µF, 500Vdc, HP Part No. 0160-0269. Add C.S.A. (Canadian Standards Association) identification label, HP Part No. 7120-8572. The Model 6824A is now C.S.A. certified for laboratory equipment.

### ERRATA:

For all instruments delivered on or after July 1, 1978, change the HP Part No. for fuseholder from 1400-0084 to fuseholder body 2100-0564 and fuseholder carcier 2100-0565. Change the HP Part No. for fuseholder nut from 2950-0038 to 2110-0569. If old fuseholder must be replaced for any reason, replace complete fuseholder and nut with new fuseholder parts. Do not replace new parts with old parts.

In the replaceable parts list, page 6-5, change Q2,13,14 to HP Part No. 1953-0485.

On page 1-3, change the Output Impedance specification to read as follows: "Less than 30 ohms from 1000Hz to 100kHz."
On page 5-6, change paragraph 5-20g to read as follows: "Values should be less than 0.3 ohm and 30 ohms respectively."

### CHANGE 21:

In the replaceable parts list, page 6-5, add C22, 56pf, 300V, Mica, HP Part No. 0140-0191, qty. 1. On the schematic, C22 is added between the base and collector of 012.

\*This change also applies to the following serial numbers:

2001A	3149	2001A	3173
	3153-3155		3178,3179
	3157		3181-3188
	3162,3163		3190
	3171	:	3198

### CHANGE 22:

In the replaceable parts list, page 6-5, change C22 (added in CHANGE 21) to 270pf, 300V, Mica, HP Part No. 0140-0210, qty..1.

tThis change also applies to the following serial numbers: 2205A 3211-3220

9-24-82

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Tet: 56 21 81/20
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